Cross-correlation method for intermediate-duration gravitational wave searches associated with gamma-ray bursts

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Several models of gamma-ray burst progenitors suggest that the gamma-ray event may be followed by gravitational wave signals of 10^3–10^4 seconds duration (possibly accompanying the so-called X-ray afterglow "plateaus"). We term these signals "intermediate duration" because they are shorter than continuous wave signals but longer than signals traditionally considered as gravitational wave bursts, and are difficult to detect with most burst and continuous wave methods. The cross-correlation technique proposed by [S. Dhurandhar et al., Phys. Rev. D **77**, 082001 (2008)], which so far has been used only on continuous wave signals, in principle unifies both burst and continuous wave (as well as matched filtering and stochastic background) methods, reducing them to different choices of which data to correlate on which time scales. Here we perform the first tuning of this cross-correlation technique to intermediate duration signals. We derive theoretical estimates of sensitivity in Gaussian noise in different limits of the cross-correlation formalism, and compare them to the performance of a prototype search code on simulated Gaussian-noise data. We estimate that the code is likely able to detect **some classes** of intermediate duration signals (such as the ones described in [A. Corsi & P. Mészáros, Astrophys. J., **702**, 1171 (2009)]) **from sources located at astrophysically-relevant distances of several tens of Mpc.**

I. INTRODUCTION

Over the last decade, the LIGO and Virgo gravitational wave (GW) detectors have carried out triggered (or targeted) GW searches in coincidence with Gamma-Ray Bursts (GRBs) and other electromagnetic transients [1–15] as well as persistent electromagnetic sources [16–28]. These searches have traditionally been optimized to detect well-modeled "chirp" signals from neutron star (NS)-NS and/or black-hole (BH)-NS binary inspirals, unmodeled short ($\lesssim 1 - 10\,\mathrm{s}$) duration bursts of GWs in association with electromagnetic transients, and persistent (continuous) GWs from nearby rotating NSs. Searches based on methods for a stochastic background have also been adapted to continuous wave targets [23, 29].

Methods targeting an as of yet largely unexplored class of "intermediate duration" GW signals have also been developed [30–32] and two so far have led to a search on real data [13, 33].[1] Intermediate duration GWs are of special interest in several astrophysical scenarios (e.g., [13, 35–43]), and their detectability over a large parameter space remains mostly unexplored compared to the more traditional inspiral, burst, or continuous wave signals.

In this work, we focus on the possibility of detecting $10^3 - 10^4\,\mathrm{s}$ duration GWs in coincidence with GRBs. Our study is motivated by the need for a data analysis technique that is optimized to probe some of the long-lived progenitor scenarios for (long and short) GRBs, such as the so-called "magnetar model". The magnetized NS (magnetar) scenario has been invoked to explain X-ray "plateaus" ($10^2 - 10^4$ s-long periods of relatively constant emission) observed in $\gtrsim 50\%$ of long, and in several short, GRB afterglows [44–51]. Gravitational collapse leading to the formation of a NS, in turn, has long been considered an observable source of GWs. In a rotating, newly born NS, non-axisymmetric instabilities such as the secular Chandrasekhar-Friedman-Schutz [CFS, 52, 53] instabilities, can yield GW emission with high efficiency [54]. If the newly born GRB-magnetar emits GWs over the plateau timescale ($\sim 10^3$ s), GW detectors such as the advanced LIGO (aLIGO) and Virgo **detectors** may be able to directly probe the source of the observed prolonged energy injection, and clarify one of the key open questions on the nature of GRB central engines [38, 55].

Detecting intermediate duration GW signals, such as the ones possibly associated with GRB plateaus, requires search techniques that can bridge the gap (both in terms of science reach and signal detection strategies) between traditional inspiral/burst searches, and continuous wave or stochastic ones. Traditional short duration inspiral and long duration continuous wave searches make use of highly sensitive coherent (and computationally limited semi-coherent) techniques that leverage accurate knowledge of the expected GW waveform (as **a** function of a set of physical parameters). Traditional burst and stochastic searches, on the other hand, assume little a priori knowledge of the signal and depend respectively on excess signal power (above the background noise) and cross-correlation of power between interferometers for detection.

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[1] Those works use "long" to refer to signals of $\mathcal{O}(10^2)$ s duration, because these durations are long compared to the $\mathcal{O}(\lesssim 1)$ s duration signals traditionally targeted in burst data analyses. The term "very long duration" signals has also been adopted to refer to GWs lasting from hours to weeks, e.g. [34]. Here, we use "intermediate" to put the discussion in the broader context, which includes the substantially longer continuous wave signals.

Here we address the problem of searching for intermediate duration, large frequency bandwidth signals by adapting the cross-correlation method of [56]. While originally developed in the context of continuous waves, the method by [56] encompasses all of the aforementioned traditional search techniques when various parameters are taken to the appropriate limits, and it shows how to make best use of the information available about each type of signal. (A Bayesian framework of similarly broad relevance was developed later in Cornish and Romano [57], but here like Dhurandhar et al. [56] we present an essentially frequentist analysis.) We correct some small errors in the original formalism of [56], and apply it for the first time to intermediate duration signals by developing a code whose performance we test on simulated data. We restrict ourselves to intermediate duration signals with large frequency bandwidth (such as the ones described in [38]), since intermediate duration narrow band signals have different astrophysical origins and are treated with adaptations of continuous wave searches (see e.g. [58]).

Our paper is organized as follows. In Sec. II we motivate the application of Dhurandhar et al.'s [56] cross-correlation technique to intermediate duration GWs. In Sec. III we describe our notation and assumptions. In Sec. IV we briefly re-derive the general statistical behavior of the cross-correlation method, discuss explicitly its limits and intermediate regimes, and show how several assumptions made in [56] need to be modified for the search of non well-modeled GW transients evolving on $10^3 - 10^4$ s timescales. In Sec. V we apply the cross-correlation technique to the model of secularly unstable GRB-magnetars described in [38], thus providing an example of applicability to astrophysically motivated waveforms of intermediate duration. Finally, in Section VI, we compare our results with other data analysis techniques that have been proposed to search for intermediate duration GW signals, and give our conclusions.

II. MOTIVATION FOR A CROSS-CORRELATION SEARCH

GWs signals are typically predicted to have strengths so close to the level of noise in the detectors that it is necessary to filter the interferometer data streams to detect the real GW events amongst spurious noise events. When the functional form of the predicted GW signal is very well known (as a function of a set of physical parameters), matched filtering with template waveforms is the optimal strategy (e.g., [59, 60]). Matched filtering involves computing the cross-correlation between the interferometer output and a template waveform, weighted inversely by the noise spectrum of the detector. The signal-to-noise ratio (SNR) is defined as the cross-correlation of the template with a particular stretch of data divided by the root-mean-squared (rms) value of the cross-correlation of the template with pure detector noise.

Usually, a family of templates spanning the possible range of parameter values (a so-called template bank) is used in real data analyses. A template bank adds to the search statistics a trial factor, which has to be taken into account when estimating the detection sensitivity. A template bank also involves more computational cost since each template must be cross-correlated with the data. While the parameters describing the search templates typically vary continuously throughout a finite range of values, a realistic template bank is composed of templates whose parameter values vary in discrete steps within the allowed range. The "mismatch" between the signal and nearest of the discrete templates causes some reduction in the expected matched filter SNR. Thus, the number of templates to be used in a search is a compromise between the maximum computational cost one can sustain, and the maximum mismatch that one is willing to tolerate (e.g., [61–64]).

When the maximum sustainable computational cost implies a mismatch such that the loss in SNR reduces the sensitivity of the search to a very limited portion of the parameter space, modifications to the matched filtering strategy toward sub-optimal techniques are mandatory. In addition, in many cases, the GW signal waveform is not known well enough for matched filtering. Indeed, even if a very finely spaced discrete template bank is used, a search may fail to detected a signal if the templates do not represent with sufficient accuracy the relevant physics. **In other words, a realistic search is affected not only by the mismatch but also by the so-called "fitting factor" [65–68], the fractional loss in SNR caused by the fact that even the best template in a family is only a "fit" to a hypothetical exact gravitational waveform. In the context of GWs from compact binaries, where numerical relativity can be used to quantify the fitting factor of phenomenological waveforms used to construct template banks for matched filter searches (e.g., [69]), it has been estimated that fitting factors $< 3\%$ are needed to achieve detection efficiencies $> 90\%$ (see e.g. [65, 70]). Indeed, matched filtering is by construction highly likely to miss a signal even for moderately bad fitting factors. On the other hand, sub-optimal (less sensitive) detection techniques are more robust against the intrinsic uncertainties in the underlying physics [71–73].**

In the case of secular bar-mode GW signals from GRB afterglow plateaus, given the uncertainties related to the physics of GRB central engines, **the derived gravitational waveforms are to be considered as simplified phenomenological models**. Thus, a more robust (when compared to matched filtering) search is necessary. A very robust approach against signal uncertainties consists of using the cross-correlation between the output of *different, non-co-located* detectors. This approach (which, differently from matched filtering, requires no a-priori knowledge of the signal waveform and its properties) is typically used for stochastic GW background

searches (e.g., [29, 74, 75]). The cross-correlation between different, non-co-located detectors, only relies on the fact that, in the presence of a GW signal, the output from distinct detectors (at the same times, after correcting for the light-travel time between detectors) should be correlated, while pure noise would remain uncorrelated. Of course, this technique also implies a poor resolution in the parameter space, and more expensive follow-ups to verify possible detections [56].

It is important to note that the cross-correlation is at the basis of two opposite search strategies: the (highly sensitive) matched filtering (cross-correlation of the data with a template), and the (very robust) "stochastic search" (cross-correlation of different detector's output). Indeed, by noticing this fundamental fact, Dhurandhar et al. 2008 [56] have provided an elegant formulation of the cross-correlation statistic for periodic GW searches such that, depending on the maximum duration over which one believes phase coherence is preserved by the signal, the statistic can be tuned to go from a "stochastic-type" search using data from distinct detectors, to the semi-coherent time-frequency methods with increasing coherent time baselines (e.g., [62]), and all the way to a fully coherent search (nearly recovering the matched filtering statistic).

Dhurandhar et al.'s formulation of the cross-correlation statistic [56] leads to a unified framework that can be used to make informed trade-offs between computational cost, sensitivity, and robustness against signal uncertainties. Studies based on the cross-correlation statistic as formulated by [56] have focused on continuous GW emission from Supernova 1987a and Scorpius X-1 [76, 77], and a number of refinements to the cross-correlation method have also been published in recent years, particularly for the treatment of spectral leakage [77, 78]. In what follows, we present a strategy tuned for the detection of *intermediate duration* ($\lesssim 10^4$ s) quasi-periodic GW signals, **and discuss its application to the case of secularly unstable GRB magnetars** (Section V).

III.   NOTATION AND ASSUMPTIONS

A.   The Short-duration Fourier Transform

The Short-time Fourier Transform (SFT) is a useful tool when examining a signal whose frequency content is evolving with time. The time-domain output of LIGO/Virgo detectors, $x(t)$, can be represented as the linear combination of a GW signal $h(t)$, and background noise $n(t)$:

$$x(t) = h(t) + n(t). \qquad (3.1)$$

The SFT of the detector output is constructed by dividing the time-series $x(t)$ into N_{SFT} segments of duration ΔT_{SFT} (generally speaking, these segments may or may not overlap), and by taking the Discrete Fourier Trans-

form (DFT) of each of these segments:

$$\tilde{x}_I[f_k] = \frac{1}{f_s} \sum_{l=0}^{N_{\mathrm{bin}}-1} x[t_l] \mathrm{e}^{-2\pi i f_k(t_l - T_I + \Delta T_{\mathrm{SFT}}/2)}, \quad (3.2)$$

where f_s is the sampling frequency (typically $f_s = 16,384\,\mathrm{Hz}$ for the LIGO detectors); $N_{\mathrm{bin}} = \Delta T_{\mathrm{SFT}} \times f_s$ is the number of frequency bins of each SFT; and f_k is the frequency corresponding to the k-th frequency bin:

$$f_k = \frac{k}{\Delta T_{\mathrm{SFT}}} \quad \text{for} \quad k = 0, ..., N_{\mathrm{bin}}/2 - 1, \quad (3.3)$$

$$f_k = \frac{(k - N_{\mathrm{bin}})}{\Delta T_{\mathrm{SFT}}} \quad \text{for} \quad k = N_{\mathrm{bin}}/2, ..., N_{\mathrm{bin}} - 1. (3.4)$$

Note that t_l in Eq. (3.2) corresponds to the l-th time sample i.e., $t_l = T_I - \Delta T_{\mathrm{SFT}}/2 + l/f_s$. For each $I = 0, 1, ...T_{\mathrm{obs}}/\Delta T_{\mathrm{SFT}}$ (where T_{obs} is the total duration of the signal) and $l = 0, 1, ..., N_{\mathrm{bin}}$, t_l spans the time interval $T_I - \Delta T_{\mathrm{SFT}}/2 \leq t_l \leq T_I + \Delta T_{\mathrm{SFT}}$. Note also that we distinguish between continuous time series $x(...)$ and their associated discretely-sampled time series $x[...]$ by using square brackets.

To reduce spectral leakage, a windowing function $w[t_l]$ is often applied to the DFT [79]:

$$\tilde{x}_I[f_k] = \sum_{l=0}^{N_{\mathrm{bin}}-1} w[t_l] x[t_l] \mathrm{e}^{-2\pi i f_k(t_l - T_I + \Delta T_{\mathrm{SFT}}/2)}. \quad (3.5)$$

For simplicity, and following [56], hereafter we neglect the window function (but discuss some of the related issues in Section IV D).

B.   The noise and its PSD

In this Section we consider the detector output in the absence of a signal. In the continuum limit of Eq. (3.1), the frequency (f) content of the detector noise can be described by its Fourier transform:

$$\tilde{n}(f) = \int_{-\infty}^{\infty} dt \ n(t) \mathrm{e}^{-2\pi i f t}. \qquad (3.6)$$

The single-sided ($f \gtrsim 0$) Power Spectral Density (PSD) of the noise, $S_n(f)$, is defined as:

$$S_n(f) := 2 \int_{-\infty}^{\infty} d\tau \ \langle n(t)n(t+\tau) \rangle \mathrm{e}^{-2\pi i f \tau}, \qquad (3.7)$$

where $\langle n(t)n(t + \tau) \rangle$ is the autocorrelation function of the noise, and the expectation value $\langle \cdot \rangle$ represents an average over an ensemble of noise realizations. The noise autocorrelation function thus forms a Fourier transform pair with its PSD. Note that hereafter we assume the noise is stationary and Gaussian (with zero mean), thus its autocorrelation function is independent of t.

From Eq. (3.6), it follows that (see also [80]):

$$\langle \tilde{n}^*(f')\tilde{n}(f) \rangle = \left\langle \int_{-\infty}^{\infty} dt' \; n^*(t') \mathrm{e}^{2\pi i f' t'} \int_{-\infty}^{\infty} dt \; n(t) \mathrm{e}^{-2\pi i f t} \right\rangle . \tag{3.8}$$

This product of independent integrals can be recast as:

$$\langle \tilde{n}^*(f')\tilde{n}(f) \rangle = \left\langle \int_{\infty}^{\infty} dt' \int_{-\infty}^{\infty} dt \; n^*(t')n(t)\mathrm{e}^{2\pi i f' t'} \mathrm{e}^{-2\pi i f t} \right\rangle . \tag{3.9}$$

Noting that real detector output implies $n^*(t) = n(t)$, and given the linearity and limited multiplicativity[2] of the expectation value, we have:

$$\langle \tilde{n}^*(f')\tilde{n}(f) \rangle = \int_{-\infty}^{\infty} dt' \int_{-\infty}^{\infty} dt \; \langle n(t')n(t) \rangle \, \mathrm{e}^{2\pi i f' t'} \mathrm{e}^{-2\pi i f t} . \tag{3.10}$$

Setting $t = t' + \tau$, yields:

$$\langle \tilde{n}^*(f')\tilde{n}(f) \rangle = \\ \int_{-\infty}^{\infty} dt' \; \mathrm{e}^{-2\pi i (f-f') t'} \int_{-\infty}^{\infty} d\tau \; \langle n(t')n(t'+\tau) \rangle \mathrm{e}^{-2\pi i f \tau} . \tag{3.11}$$

Then, using Eq. (3.7), we replace the integral over $d\tau$ with the PSD,

$$\langle \tilde{n}^*(f')\tilde{n}(f) \rangle = \frac{S_n(f)}{2} \int_{-\infty}^{\infty} dt' \; \mathrm{e}^{-2\pi i (f-f') t'} . \tag{3.12}$$

The remaining integral over dt' is simply a delta function,

$$\langle \tilde{n}^*(f')\tilde{n}(f) \rangle = \frac{1}{2}\delta(f - f')S_n(f), \tag{3.13}$$

and using the finite time approximation of the delta function:

$$\delta_{\Delta T_{\mathrm{SFT}}}(f) = \frac{\sin(\pi f \Delta T_{\mathrm{SFT}})}{\pi f}, \tag{3.14}$$

which reduces to ΔT_{SFT} in the limit of $f \to 0$, we can relate the variance of the Fourier transformed detector output to the PSD:

$$\langle |\tilde{n}_I[f_k]|^2 \rangle \approx \frac{\Delta T_{\mathrm{SFT}}}{2} S_n[f_k]. \tag{3.15}$$

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[2] The expectation value $\langle XY \rangle$ of random variables X, Y is multiplicative if $\mathrm{Cov}(X,Y) = 0$. That is, only if X and Y are statistically independent.

C. Short-duration Fourier Transform of the signal

We make the hypothesis that the GW signal $h(t)$ is quasi-periodic (by taking a sufficiently small time interval the signal in such an interval can be considered monochromatic), and assume that its time-frequency evolution is described with sufficient physical accuracy, for a time interval of T_{coh}, via some known function of a given set of parameters (although this function may not have a closed form expression). By definition, this "coherence timescale" is less than or equal to the total observation time T_{obs} over which the signal is expected to last (e.g. $T_{\mathrm{coh}} \lesssim T_{\mathrm{obs}} \lesssim 10^4$ s for the type of signals of interest in the context of GRB afterglow plateaus).

Since the signal is quasi-periodic, we can define an SFT baseline $\Delta T_{\mathrm{SFT}} \leq T_{\mathrm{coh}}$ such that, within the baseline, all of the signal power is concentrated in a single SFT bin. More specifically, around each time T_I we can approximate the signal received by the detector in the time interval $T_I - \frac{\Delta T_{\mathrm{SFT}}}{2} \lesssim t \lesssim T_I + \frac{\Delta T_{\mathrm{SFT}}}{2}$, as:

$$h(t) \approx h_0(T_I)\mathcal{A}_+ F_+ \cos(\Phi(T_I) + 2\pi f(T_I)(t - T_I)) + \\ h_0(T_I)\mathcal{A}_\times F_\times \sin(\Phi(T_I) + 2\pi f(T_I)(t - T_I)), \tag{3.16}$$

where $\mathcal{A}_+$, $\mathcal{A}_\times$ are amplitude factors dependent on the physical system's inclination angle ι (for on-axis GRBs, ι is the angle between the jet axis and the line of sight):

$$\mathcal{A}_+ = \frac{1 + \cos^2 \iota}{2}, \tag{3.17}$$

$$\mathcal{A}_\times = \cos \iota, \tag{3.18}$$

and F_+, $F_\times$ are the antenna factors that quantify the detector's sensitivity to each polarization state. Note that for triggered searches targeting GRBs (as is the case in Sec. V), the line of sight is expected to be nearly aligned with the jet axis,[3] thus $\iota \approx 0$ and $\mathcal{A}_+ \approx \mathcal{A}_\times \approx 1$.

In order for the approximation in Eq. (3.16) to be valid, the following conditions should be satisfied:

1. $T_{\mathrm{obs}} \lesssim 10^4$ s so that, for a given GW detector, F_+ and $F_\times$ can be treated as constants as a function of time (see e.g. [56]).

2. If $\dot{f}(t)$ is the time derivative of the signal frequency at a given time t, then the effects of $\dot{f}(t)$ on the signal phase should be negligible during the time interval ΔT_{SFT}. Using the quarter-cycle criterion, this leads to $2\pi|\dot{f}(T_I)|\left(\frac{\Delta T_{\mathrm{SFT}}}{2}\right)^2 < \frac{\pi}{2}$. Thus, $\Delta T_{\mathrm{SFT}} < 1/\sqrt{|\dot{f}(T_I)|}$.

3. ΔT_{SFT} is small enough that $h_0(t) \approx h_0(T_I)$ (constant amplitude approximation) in the interval

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[3] That is, the line of sight is within the jet-opening angle, which is expected to be of the order $5 - 20 \deg$ for long GRBs [81, 82].

$T_I - \Delta T_{\text{SFT}}/2 \lesssim t \lesssim T_I + \Delta T_{\text{SFT}}/2$. We consider this condition satisfied if $\left|\dot{h}_0(T_I)\right| \Delta T_{\text{SFT}}/h_0(T_I) \lesssim 10\%$, based on typical LIGO amplitude calibration errors ($\sim 10\%$ [83]; thus, any change of signal amplitude below 10% is not expected to significantly affect the goodness of this approximation).

In addition, hereafter we assume that ΔT_{SFT} is large enough that the corresponding frequency resolution, $(\Delta T_{\text{SFT}})^{-1}$, still enables one to track the time-frequency evolution of the signal.

Using Eq. (3.2), we can calculate the DFT of the signal in Eq. (3.16) (see also Eq. (2.25) in [56]):

$$\tilde{h}_I[f_k] = h_0(T_I)e^{i\pi f_{k,I}\Delta T_{\text{SFT}}} \times$$
$$[e^{i\Phi(T_I)}\frac{\mathcal{A}_+ F_{+,I} - i\mathcal{A}_\times F_{\times,I}}{2}\delta_{\Delta T_{\text{SFT}}}(f_k - f_{k,I})+$$
$$e^{-i\Phi(T_I)}\frac{\mathcal{A}_+ F_{+,I} + i\mathcal{A}_\times F_{\times,I}}{2}\delta_{\Delta T_{\text{SFT}}}(f_k + f_{k,I})], \quad (3.19)$$

or, equivalently,

$$\tilde{h}_I[f_k] = \frac{\sqrt{\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2}}{2}h_0(T_I)e^{i\pi f_{k,I}\Delta T_{\text{SFT}}}$$
$$\times \left[e^{i\Phi(T_I)}e^{i\varphi_I}\delta_{\Delta T_{\text{SFT}}}(f_k - f_{k,I})+\right.$$
$$\left. e^{-i\Phi(T_I)}e^{-i\varphi_I}\delta_{\Delta T_{\text{SFT}}}(f_k + f_{k,I})\right], \quad (3.20)$$

where we have set:

$$\mathcal{A}_+ F_{+,I} \pm i\mathcal{A}_\times F_{\times,I} = \sqrt{\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2}e^{\mp i\varphi_I},$$
$$(3.21)$$

and

$$\varphi_I = \arctan(-\mathcal{A}_\times F_{\times,I}/\mathcal{A}_+ F_{+,I}). \quad (3.22)$$

Note that, while in our limit of intermediate duration GW signals the antenna response from one detector can be considered constant over the observed duration of the signal, for the multiple detector case the antenna responses refer to the specific GW detector from whose output the I-th SFT is taken.

IV. THE CROSS-CORRELATION STATISTIC

Following [56], we define the raw cross-correlation statistic as:

$$\mathcal{Y}_{IJ} = \frac{\tilde{x}_I^*[f_{k,I}]\tilde{x}_J[f_{k',J}]}{\Delta T_{SFT}^2}, \quad (4.1)$$

where the frequency $f_{k,I}$ is the frequency at which all of the signal power is concentrated during the Ith time interval (see Eq. (3.19)), and is related to the frequency

$f_{k',J}$ at which all of the signal power is concentrated during the Jth time interval via the relation:

$$f_{k',J} = f_{k,I} - \Delta f_{IJ}. \quad (4.2)$$

In the above relation, Δf_{IJ} is the frequency difference predicted by the model's time-frequency evolution (in this analysis the signal time-frequency evolution is assumed to be known to some level of accuracy; see Section III C). Note that, because for any I-th SFT the associated frequency bin k is fixed by the model's predictions, we omit the indexes k, k' from $\mathcal{Y}_{IJ}$ for simplicity.

For a signal embedded in stationary Gaussian noise with zero mean, the $\{\mathcal{Y}_{IJ}\}$ are themselves random variables with mean and variance given by

$$\mu_{IJ} = h_0(T_I)h_0(T_J)\tilde{\mathcal{G}}_{IJ}, \quad (4.3)$$
$$\sigma_{IJ}^2 = \frac{1}{4\Delta T_{\text{SFT}}^2}S_n[f_{k,I}]S_n[f_{k',J}], \quad (4.4)$$

where we have used Eqs. (3.15) and (3.19), and the fact that:

$$\tilde{h}_I^*[f_k]\tilde{h}_J[f_k + \Delta f_{IJ}] = h_0(T_I)h_0(T_J)\tilde{\mathcal{G}}_{IJ}\delta_{\Delta T_{\text{SFT}}}^2(f_k - f_{k,I}).$$
$$(4.5)$$

In the above equations, $\tilde{\mathcal{G}}_{IJ}$ is the signal cross-correlation function, defined here as

$$\tilde{\mathcal{G}}_{IJ} = \frac{\sqrt{\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2}}{2}\frac{\sqrt{\mathcal{A}_+^2 F_{+,J}^2 + \mathcal{A}_\times^2 F_{\times,J}^2}}{2}e^{-i\Delta\theta_{IJ}},$$
$$(4.6)$$

with $\Delta\theta_{IJ} = \theta_I - \theta_J = \pi\Delta f_{IJ}\Delta T_{\text{SFT}} + \Delta\Phi_{IJ} + \Delta\varphi_{IJ}$. In general, the subscripts (I), (J) in the antenna responses refer to the specific GW detector from whose output the I-th (or J-th) SFT is taken. Indeed, in the definition of the $\{\mathcal{Y}_{IJ}\}$, there is total freedom to correlate pairs from one single detector or from an arbitrary number of detectors.

Note that the $e^{-i\pi\Delta f_{IJ}\Delta T_{\text{SFT}}}$ term that arises from $\Delta\theta_{IJ}$ in Eq. (4.6) is absent from the definition of the signal-cross-correlation function given in [56]. This discrepancy was first noted in [76], and is discussed there in detail. This term proves essential to properly tracking the frequency evolution of a given signal across SFTs, so we call attention to it here.

When cross-correlation pairs are only taken from the output of a *single* detector over timescales of $T_{\text{obs}} \lesssim 10^4$ s, then $F_{+,\times,I} = F_{+,\times,J} = F_{+,\times}$. This simplifies Eq. (4.6) considerably:

$$\tilde{\mathcal{G}}_{IJ}^{\text{1D}} = \frac{\mathcal{A}_+^2 F_+^2 + \mathcal{A}_\times^2 F_\times^2}{4}e^{-i\pi\Delta f_{IJ}\Delta T_{\text{SFT}}}e^{-i\Delta\Phi_{IJ}}. \quad (4.7)$$

For two or more detectors, such as LIGO Hanford (H) and LIGO Livingston (L), the indexes I and J are free to range over SFTs from *either* detector, and so the above simplification does not generally apply (even if the antenna factors for each detector are approximately constant within the considered time interval).

Following [56], our detection statistic is then constructed as a weighted sum of the $\mathcal{Y}_{IJ}$

$$\rho = \sum_{IJ}(u_{IJ}\mathcal{Y}_{IJ} + u_{IJ}^*\mathcal{Y}_{IJ}^*), \qquad (4.8)$$

with nearly optimal weights[4]

$$u_{IJ} = \frac{\tilde{\mathcal{G}}_{IJ}^*}{\sigma_{IJ}^2}. \qquad (4.9)$$

For stationary Gaussian distributed white noise (see Eq. 4.4), σ_{IJ} does not depend on frequency nor on time, but it might still depend on the detector. Thus:

$$\sigma_{IJ}^2 = \frac{1}{4\Delta T_{\mathrm{SFT}}^2}S_n^2, \qquad (4.10)$$

for IJ pairs from a single detector (or identical detectors), or:

$$\sigma_{IJ}^2 = \frac{1}{4\Delta T_{\mathrm{SFT}}^2}S_n^H S_n^L, \qquad (4.11)$$

for e.g. a LIGO Hanford-Livingston IJ pair. Thus, using the above equations and Eq. (4.6), we have in general:

$$u_{IJ} = \frac{\sqrt{(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)(\mathcal{A}_+^2 F_{+,J}^2 + \mathcal{A}_\times^2 F_{\times,J}^2)}}{\Delta T_{\mathrm{SFT}}^{-2}e^{-i\Delta\theta_{IJ}}S_n[f_{k,I}]S_n[f_{k,J}]}, \qquad (4.12)$$

where, again, the antenna responses and detector's noise refer to the specific GW detector from whose output the I-th (or J-th) SFT is taken.

As we describe in more detail in what follows, the mean and variance of ρ, as well its statistical distribution, depend on the choice of which SFT pairs are cross-correlated. Because of the freedom in choosing which data-segment pairs to correlate, we can naturally consider one single detector or an arbitrary number of detectors (with no need to modify our statistic), and we can work in one of the following limits [56]:

1. We can choose to correlate only data segments taken from distinct detectors at the same times (after correcting for the light travel time between different detectors; Section IV A). This limit is analogous in spirit to the methods of stochastic GW searches, such as [84–87], and we hence refer to it as the "stochastic limit". In this case, the computational cost of the search is small and the search is very robust against signal uncertainties. But the sensitivity is the poorest, as is the resolution in parameter space.

2. At the other extreme, we can correlate all possible SFT segments (Section IV B): This (nearly) corresponds to a full matched filter statistic described for coalescing compact binaries and continuous waves in e.g. [60, 62, 88, 89]. The parameter space resolution becomes very fine and while this is ideally the most sensitive method, is it also the most computationally expensive (prohibitive for wide parameter space searches) and the least robust against signal uncertainties.

3. In intermediate regimes, we can correlate data segments separated by a maximum coherence time $T_{\mathrm{coh}} \lesssim T_{\mathrm{obs}}$ (Section IV C). This "semi-coherent" approach is similar to several methods used for continuous waves [71, 90–93] (though on signal timescales much longer than what considered in this work). Because in this limit the sensitivity and robustness of the search can be tuned to the expected accuracy of a given model, this is the regime of greatest interest to us.

4. Finally, one can consider all pairs except self-correlations. This was the main focus of the analysis presented in [56] (see their Section IV). Here, we do not focus on this limit because we consider it as a more special case of the ones above (with no particular advantages for the detection of the type of signals considered in our study and with some complications added to the statistical properties of ρ). However, in what follows, we do discuss the main differences of (1)-(3) above with respect to this case (see also Section IV of [56]).

In discussing the above limits, it is useful to note that we can re-write Eq. (4.8) in terms of Eq. (4.1) as:

$$\rho = \frac{1}{\Delta T_{\mathrm{SFT}}^2}\sum_{IJ}u_{IJ}\tilde{x}_I^*[f_{k,I}]\tilde{x}_J[f_{k',J}] + u_{IJ}^*\tilde{x}_I[f_{k,I}]\tilde{x}_J^*[f_{k',J}], \qquad (4.13)$$

which is equivalent to:

$$\rho = \frac{2}{\Delta T_{\mathrm{SFT}}^2}\sum_{IJ}\Re\left\{u_{IJ}\tilde{x}_I^*[f_{k,I}]\tilde{x}_J[f_{k',J}]\right\}. \qquad (4.14)$$

A. Stochastic limit (independent pairs only)

Consider the output of two different detectors, $\tilde{x}^H$ and $\tilde{x}^L$. Each detector's output can be divided into $T_{\mathrm{obs}}/\Delta T_{\mathrm{SFT}} = N_{\mathrm{SFT}}$ segments. Of the $(2N_{\mathrm{SFT}})^2$ possible SFT pairs that can contribute to ρ we correlate only pairs of SFTs from *different* detectors at the same time (after correcting for the light travel time between detectors), so that $N_{\mathrm{pairs}} = N_{\mathrm{SFT}}$. In this limit, Eq. (4.13) becomes:

$$\rho = 2\sum_I \Re\left\{u_{II}\mathcal{Y}_{II}\right\}, \qquad (4.15)$$

---

[4] Strictly speaking, these weights are only optimal when self-pairs are excluded, as in [56]. For sufficiently small amplitude signals, these weights remain optimal, to first order, even when self-pairs are considered. For situations where this may not be the case, we refer the reader to the discussion in the Appendix of [56].

where the weights are described by e.g. Eq. (4.11). Written explicitly, this becomes

$$\rho = \frac{2}{\Delta T_{\mathrm{SFT}}^2} \sum_I \Re \left\{ u_{II} \tilde{x}_I^{*\mathrm{H}}[f_{k,I}] \tilde{x}_I^{\mathrm{L}}[f_{k',I}] \right\}, \qquad (4.16)$$

i.e., a weighted sum of completely independent random variables that are each the product of two Gaussian variables with mean and variance given by Eqs. (4.3) and (4.4). Thus, ρ converges to a Gaussian distribution (by the Central Limit Theorem) with mean (see Eqs. (4.3), (4.6), (4.12), and [56]) and variance (see also Eq. (4.4) and Dhurandhar et al. [56]):

$$\mu_\rho = (\mathcal{A}_+^2 F_{+,H}^2 + \mathcal{A}_\times^2 F_{\times,H}^2)(\mathcal{A}_+^2 F_{+,L}^2 + \mathcal{A}_\times^2 F_{\times,L}^2) \frac{\Delta T_{\mathrm{SFT}}^2}{2} \sum_I \frac{h_0^2(T_I)}{S_n^H[f_{k,I}] S_n^L[f_{k,I}]}, \qquad (4.17)$$

$$\sigma_\rho^2 = (\mathcal{A}_+^2 F_{+,H}^2 + \mathcal{A}_\times^2 F_{\times,H}^2)(\mathcal{A}_+^2 F_{+,L}^2 + \mathcal{A}_\times^2 F_{\times,L}^2) \frac{\Delta T_{\mathrm{SFT}}^2}{2} \sum_I \frac{1}{S_n^H[f_{k,I}] S_n^L[f_{k,I}]}. \qquad (4.18)$$

The detection threshold is easily derived in terms of the Cumulative Distribution Function (CDF) of a normal distribution,

$$\mathcal{F}_\mathcal{N}(\rho) = \frac{1}{2} \left[2 - \mathrm{erfc} \left(\frac{\rho - \mu_\rho}{\sigma_\rho \sqrt{2}} \right) \right], \qquad (4.19)$$

and its inverse (see also [56]), where erfc is the complementary error function. For a False Alarm Probability (FAP) α, the associated threshold is simply $1 - \alpha = \mathcal{F}_\mathcal{N}(\rho_{\mathrm{th}})$, thus:

$$\rho_{\mathrm{th}} = \sqrt{2} \sigma_\rho \mathrm{erfc}^{-1}(2\alpha), \qquad (4.20)$$

where we have used the fact that the background distribution is considered in the absence of a signal ($\mu_\rho = 0$). When a signal is present, the detection probability γ, or, equivalently, the False Dismissal Probability (FDP) $1 - \gamma$, is given by $\gamma = \mathcal{F}_\mathcal{N}(\rho_{\mathrm{th}})$, i.e.:

$$\gamma = \frac{1}{2} \mathrm{erfc} \left(\frac{\rho_{\mathrm{th}} - \mu_\rho}{\sigma_\rho \sqrt{2}} \right). \qquad (4.21)$$

Thus, the detectability condition reads:

$$\frac{\mu_\rho}{\sigma_\rho} \gtrsim \sqrt{2} \mathcal{S}, \qquad (4.22)$$

where $\mathcal{S} = \mathrm{erfc}^{-1}(2\alpha) - \mathrm{erfc}^{-1}(2\gamma)$. In the case of white Gaussian noise, using Eqs. (4.17)–(4.18), the detectability condition implies:

$$h_{\mathrm{rms}} \gtrsim \frac{\sqrt{2} \mathcal{S}^{1/2} \Delta T_{\mathrm{SFT}}^{-1/2} N_{\mathrm{SFT}}^{-1/4} (S_n^H S_n^L)^{1/4}}{[(\mathcal{A}_+^2 F_{+,H}^2 + \mathcal{A}_\times^2 F_{\times,H}^2)(\mathcal{A}_+^2 F_{+,L}^2 + \mathcal{A}_\times^2 F_{\times,L}^2)]^{1/4}}, \qquad (4.23)$$

which generalizes Eq. (4.15) in Dhurandhar et al. [56] to the case of a non-constant signal amplitude for which (see also Eq. (3.16)):

$$h_{\mathrm{rms}} = \sqrt{\langle h_0^2(T_I) \rangle_I} = \sqrt{\frac{\sum_I h_0^2(T_I)}{N_{\mathrm{SFT}}}}. \qquad (4.24)$$
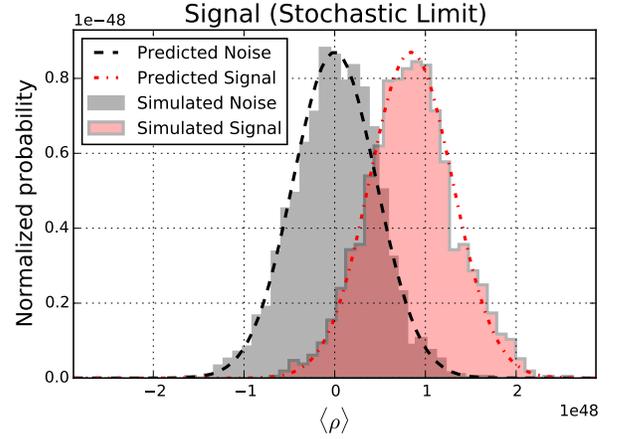
FIG. 1. Comparison between the simulated and predicted distribution of ρ in the stochastic limit, for 2048 s of simulated white Gaussian noise sampled at a rate of $f_s = 2048$ Hz, from two detector's outputs $x^H[t]$, $x^L[t]$. We have used an SFT baseline of $\Delta T_{\mathrm{SFT}} = 2$ s and, for simplicity, we assumed two idealized, co-located, and optimally oriented detectors with aLIGO-equivalent PSDs $S_n \approx 1.75 \times 10^{-47}$ Hz^{-1}, see Fig. 2. The simulated signal is a line of constant frequency $f_0 = 128$ Hz and constant amplitude $h_0 \approx 10^{-24}$.

In Fig. 1 we show the distribution of ρ in the absence of a signal for simulated Gaussian white noise, and in the presence of a GW signal of constant amplitude h_0 and constant frequency f_0. (A signal with constant frequency represents the simplest time-frequency evolution to which the technique here presented can be applied and is particularly useful for illustrative purposes.)

We stress that the independence of the pairs that are added in ρ is essential for the validity of the conclusion regarding the Gaussianity of ρ, and for the validity of Eqs. (4.23)-(4.24). While pairs are truly independent in the stochastic limit analyzed in this Section, this is not strictly true for the
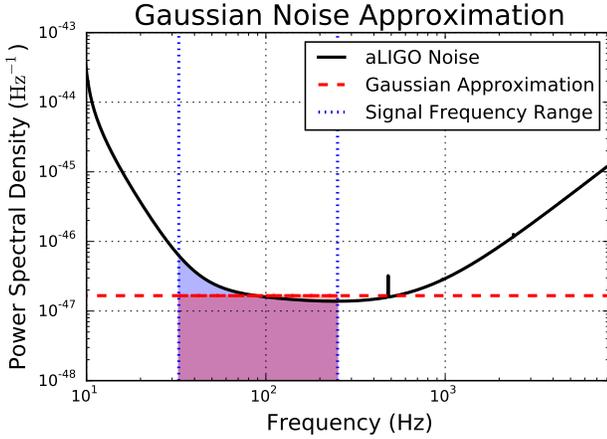
FIG. 2. Method for generating Gaussian-distributed noise with aLIGO PSD. A frequency range is defined with respect to the maximum and minimum value of an injected signal's frequency (blue dotted lines). The constant Gaussian PSD S_n (red dashed line) is calculated so that the area underneath it (shaded red) is equal to the area underneath the aLIGO PSD (shaded blue). For signals of constant frequency f_0 (e.g. Figs. 1, 3, and 4), the red area is taken between $0.9f_0$ and $1.1f_0$.

combination of pairs considered in Section IV of [56] (ρ includes all possible pairs but self ones - see also case 4 in Section IV) and in the Appendix of [56] (ρ includes all possible SFT pairs - see also case 2 in Section IV). In these cases, ρ is a sum of products that are *not* all independent. Thus, while the expressions for the mean and variance of ρ presented in Section IV of [56] (or, equivalently, Eqs. (4.17) and (4.18) here) remain valid, we caution the reader that the *lack of independence affects the shape of the background distribution*, and in some limits, results in a distribution that *cannot be reduced to a Gaussian. Thus, the detection threshold needs to be modified accordingly.* Some brief discussion of these corrections to [56] is also presented in Appendix B of [77]. In what follows, our in depth discussion of cases 2–3 (Section IV) shows explicitly that the corrections to [56] are crucial for the detection of the family of intermediate duration GW signals that we target in this analysis.

B. Matched filter limit (all pairs)

In this limit, we choose to correlated all possible SFT segments (from one or multiple detectors). Starting from Eq. (4.14), we replace the weights with their explicit form given by Eq. (4.12),

$$\rho = 2\Re\left[\sum_{I,J}^{N_{\text{pairs}}} \frac{\sqrt{(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)}}{S_n[f_{k,I}]} \frac{\sqrt{(\mathcal{A}_+^2 F_{+,J}^2 + \mathcal{A}_\times^2 F_{\times,J}^2)}}{S_n[f_{k,J}]} \tilde{x}_I^*[f_{k,I}]\tilde{x}_J[f_{k,J}]e^{i\Delta\theta_{IJ}}\right], \quad (4.25)$$

where $N_{\text{pairs}} = N_{\text{SFT}}^2$ and $N_{\text{SFT}} = N_{\text{det}}T_{\text{obs}}/\Delta T_{\text{SFT}}$, with N_{det} being the number of detectors from which data are taken. Under the change of variable

$$\tilde{x}_I'[f_{k,I}] = \frac{\sqrt{(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)}}{S_n[f_{k,I}]}\tilde{x}_I[f_{k,I}]e^{-i\theta_I}, \quad (4.26)$$

Eq. (4.25) simplifies to:

$$\rho = 2\left\{\sum_I^{N_{\text{SFT}}} |\tilde{x}_I'[f_{k,I}]|^2 + 2\sum_{I>J}^{N_{\text{SFT}}} \Re\left[\tilde{x}_I'^*[f_{k,I}]\tilde{x}_J'[f_{k',J}]\right]\right\}. \quad (4.27)$$

It then follows that,

$$\rho = 2\left|\sum_I^{N_{\text{SFT}}} \tilde{x}_I'[f_{k,I}]\right|^2. \quad (4.28)$$

Or alternatively,

$$\rho = 2\left[\left(\sum_I^{N_{\text{SFT}}} \Re(\tilde{x}_I'[f_{k,I}])\right)^2 + \left(\sum_I^{N_{\text{SFT}}} \Im(\tilde{x}_I'[f_{k,I}])\right)^2\right]. \quad (4.29)$$

For stationary Gaussian noise with zero mean, $\tilde{x}_I[f_k]$ follows a complex normal distribution. We note that the scaling and complex rotation applied in Eq. (4.26) have no effect on the shape of the distribution of the $\tilde{x}_I'$ when compared to the $\tilde{x}_I$ (but they do change the mean and variance of the distribution). Thus, the real and imaginary parts of $\tilde{x}_I'$ are still Gaussian distributed as the $\tilde{x}_I$, and so are their sums. Indeed, in the absence of a signal, the sums of the real and imaginary parts of the $\tilde{x}_I'$ are Gaussian variables with zero mean and variance (see Eq. (3.15)):

$$\sigma_\Sigma^2 = \sum_I^{N_{\text{SFT}}} \left[\frac{\Delta T_{\text{SFT}}(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)}{4S_n[f_{k,I}]}\right]. \quad (4.30)$$

So we can re-write the expression for ρ as:

$$\rho = C_\chi \times \left[\left(\frac{\sum_I^{N_{\text{SFT}}} \Re(\tilde{x}_I'[f_{k,I}])}{\sigma_\Sigma}\right)^2 + \left(\frac{\sum_I^{N_{\text{SFT}}} \Im(\tilde{x}_I'[f_{k,I}])}{\sigma_\Sigma}\right)^2\right], \quad (4.31)$$
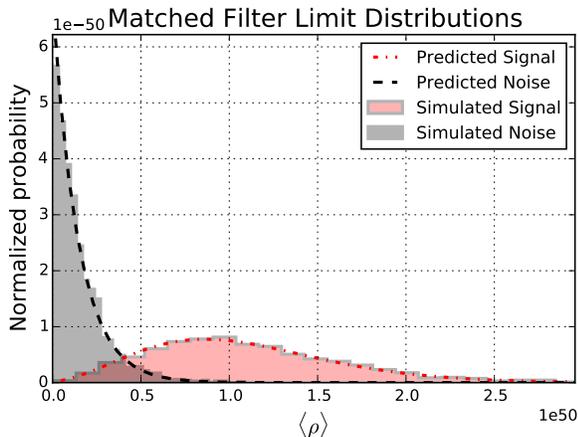
FIG. 3. We simulate 2048 s of white Gaussian noise for a single optimally-oriented detector with aLIGO-equivalent noise PSD given by $S_n \approx 1.75 \times 10^{-47}$ Hz^{-1}. We used a sampling frequency of $f_s = 2048$ Hz and an SFT baseline of $\Delta T_{\mathrm{SFT}} = 2$ s. All possible pairs are included in the cross-correlation statistic ρ which is thus distributed as a scaled χ^2 distribution with 2 degrees of freedom. The simulated signal was a line of constant frequency $f_0 = 128$ Hz and constant amplitude $h_0 \approx 3.30 \times 10^{-25}$.

which is the sum of the squares of two normally distributed variables, scaled by a factor:

$$C_\chi = 2\sigma_\Sigma^2 = \sum_I^{N_{\mathrm{SFT}}} \left[\frac{\Delta T_{\mathrm{SFT}}(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)}{2 S_n[f_{k,I}]} \right]. \tag{4.32}$$

Thus, the resulting ρ statistic is distributed as a χ^2 with 2 degrees of freedom (Fig. 3; see also [77]).

Continuing from Eq. (4.31), in the absence of a signal, the variance of ρ is simply,

$$\sigma_\rho^2 = 4C_\chi = 2 \sum_I^{N_{\mathrm{SFT}}} \left[\frac{\Delta T_{\mathrm{SFT}}(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)}{S_n[f_{k,I}]} \right]. \tag{4.33}$$

In the presence of a signal, the distribution of ρ in Eq. (4.31) becomes a non-central χ^2 with two degrees of freedom, $\chi_{\mathrm{NC}}^2(2; \lambda)$, of mean:

$$\mu_\rho = C_\chi(2 + \lambda). \tag{4.34}$$

The non-centrality parameter can be derived using the above relation, and noting that μ_ρ can be easily calculated using Eqs. (4.5), (4.6), and (4.25). This yields to (see also Eq. (4.24) and Fig. 3):

$$\lambda = \sum_I^{N_{\mathrm{SFT}}} h_0^2(T_I) \left[\frac{\Delta T_{\mathrm{SFT}}(\mathcal{A}_+^2 F_{+,I}^2 + \mathcal{A}_\times^2 F_{\times,I}^2)}{S_n[f_{k,I}]} \right]. \tag{4.35}$$

Note that in this limit the number of SFT pairs only affects the variance (and mean) of the two Gaussian variables $\sum_I^{N_{\mathrm{SFT}}} \Re(\tilde{x}_I'[f_{k,I}])$ and $\sum_I^{N_{\mathrm{SFT}}} \Im(\tilde{x}_I'[f_{k,I}])$. It does not affect the number of degrees of freedom in ρ, which remains two independently of the number of SFTs. Thus, as N_{SFT} increases, the distribution of ρ *does not approach a Gaussian. This is a critical distinction to make, since it changes the (false alarm and false dismissal) thresholds of ρ significantly from the ones that were adopted in the appendix of [56], where a Gaussian distribution was incorrectly assumed for ρ.*

In the case in which all pairs come from a single detector (or from co-located, equally oriented detectors, with identical S_n), the variance of ρ simplifies substantially to:

$$\sigma_\rho^2 = 4C_\chi = 2T_{\mathrm{obs}} \left[\frac{(\mathcal{A}_+^2 F_+^2 + \mathcal{A}_\times^2 F_\times^2)}{S_n} \right], \tag{4.36}$$

where we have used $T_{\mathrm{obs}} = N_{\mathrm{SFT}} \Delta T_{\mathrm{SFT}}$. **The non-centrality parameter likewise simplifies, yielding,**

$$\lambda = \frac{h_{\mathrm{rms}}^2 T_{\mathrm{obs}}(\mathcal{A}_+^2 F_+^2 + \mathcal{A}_\times^2 F_\times^2)}{S_n}, \tag{4.37}$$

where we have used Eq. (4.24).

In either case, the corresponding detection threshold for a given false alarm and detection rate is now substantially different than in the stochastic limit:

$$\rho_{\mathrm{th}} = C_\chi \mathcal{F}_\chi^{-1}(1 - \alpha; 2), \tag{4.38}$$

$$\gamma = \mathcal{F}_{\mathrm{NC}\chi}(\rho_{\mathrm{th}}/C_\chi; 2, \lambda). \tag{4.39}$$

The CDF for the $\chi^2(2)$ is known in closed form (and even invertible), while the CDF for the non-central case can be calculated numerically, with results as shown in Fig. 5.

In this limit, the sensitivity approaches that of matched filtering. However there is one significant error in the description in [56]: the limit approached is that of filtering with an unknown overall phase constant, which is commonly handled by summing the squares of two matched filters a quarter cycle out of phase with each other—e.g., [61]. Hence the resulting statistic is distributed as a χ^2 with 2 degrees of freedom rather than a Gaussian. Under idealized circumstances, this reduces the sensitivity by approximately 13% with respect to a Gaussian distribution (with FAP=0.1% and FDP=50%).

C. Semi-coherent regime

As discussed in Section II, the semi-coherent regime is the most relevant for an astrophysically motivated search where the expected GW signal is known to limited accuracy. In this regime, the total observation time T_{obs} is broken up into N_{coh} coherent segments, each of duration T_{coh}. The coherence time (T_{coh}) is once again defined as the length of time wherein the signal is expected
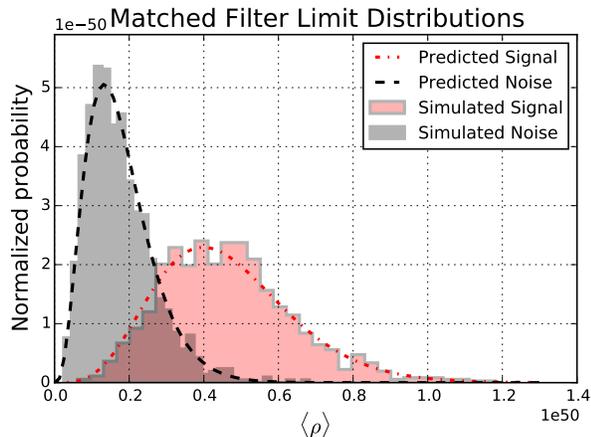
FIG. 4. Comparison between the simulated and predicted distribution of ρ in the semi-coherent limit, for 1024 s of simulated white Gaussian noise sampled at a rate of $f_s = 2048$ Hz, from one detector's output $x(t)$. We have used an SFT baseline of $\Delta T_{\mathrm{SFT}} = 2$ s and we assumed an optimally oriented detector with PSD $S_n \approx 1.91 \times 10^{-47}$ Hz^{-1}. The coherence time is $T_{\mathrm{coh}} = 256$ s for a total of $N_{\mathrm{coh}} = 4$ coherent segments. The simulated signal was a line of constant frequency $f_0 = 128$ Hz and constant amplitude $h_0 \approx 8.47 \times 10^{-25}$.

to maintain phase coherence (and therefore good agreement) with the model predictions. All possible SFT-pairs within each coherent time segment are cross-correlated, and the results for each segment are then combined incoherently.[5]

In order for the resulting sum of $\chi^2(2)$ distributed variables to add to a $\chi^2(2N_{\mathrm{coh}})$ distributed detection statistic, it is essential that all coherent segments have identical scale factors.[6] This condition is satisfied for a detector network of arbitrary size only if the detectors have **similar antenna factors for the given sky location of the event**, and each detector has (stationary) white Gaussian noise (although the frequency independent S_n of the detectors need not be identical). In the case of colored noise, the scale factors will vary between coherence segments (since the frequency of the signal is evolving with time, which causes $S_n[f_{k,I}]$ to change from segment to segment). **Thus, in the presence of colored noise, whitening the data over the signal bandwidth prior to analysis is desirable.**

Changes in the antenna factors F_+, $F_\times$ over the

duration of a signal in a non-idealized search i.e., deviations from assumption 1 in Section III C, can also affect the statistic. For the GRB X-ray plateaus of interest to Section V, $>50\%$ of events with sufficiently shallow plateau decays[7] have plateau durations $\lesssim 10^4$ s [94]. For circularly-polarized signals of this duration, we tentatively estimate that time-varying antenna factors will cause fluctuations of $\approx 15\%$ in amplitude sensitivity, comparable to LIGO amplitude calibration uncertainties [83]. We leave to future work a more in depth examination of deviations from this assumption.

When all coherent segments have identical scale factors, ρ is an incoherent sum of N_{coh} *independent* variables, each distributed as a scaled $\chi^2(2)$ distribution. The scale parameter for each coherent segment is given by Eq. (4.32) but now with $N_{\mathrm{SFT}} = T_{\mathrm{coh}}/\Delta T_{\mathrm{SFT}}$, so that:

$$C_\chi^{\mathrm{SC}} = \frac{C_\chi}{N_{\mathrm{coh}}}. \qquad (4.40)$$

The variance of the semi-coherent ρ then reads:

$$\sigma_{\rho,\mathrm{SC}}^2 = 2C_\chi^{\mathrm{SC}}(2N_{\mathrm{coh}}) = 4C_\chi, \qquad (4.41)$$

which is identical to the variance in matched-filter limit, see Eq. (4.33).

When a signal is present, the non-centrality parameter for each coherent segment will, in general, vary from one semi-coherent chunk to the other due to the time-varying amplitude of the signal in Eq. (4.35). But, since the total λ for the semi-coherent regime is additive across *all* coherent segments, the *total* non-centrality parameter for the semi-coherent ρ likewise remains unchanged from the matched filter limit. The resulting distribution thus has mean:

$$\mu_{\rho,\mathrm{SC}} = C_\chi^{\mathrm{SC}}(2N_{\mathrm{coh}} + \lambda). \qquad (4.42)$$

The above Equation reduces to (4.34) in the limit of $N_{\mathrm{coh}} = 1$ (matched-filter limit). The detection threshold for a given signal will differ from the matched filter limit due to the higher number of degrees of freedom of the χ^2 distribution of the semi-coherent ρ, and can be calculated numerically as shown in Fig. 5 (along with other limits).

In the limit of large N_{coh}, the $\chi^2(2N_{\mathrm{coh}})$ distribution tends toward a Gaussian. Continuous wave searches using this cross-correlation technique, e.g. [77] can have N_{coh} of order 10^4, and hence can set their thresholds based on Gaussian statistics as assumed by [56]. However, *searches for intermediate duration GW signals*

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[5] An alternative, but equivalent, description is to define a "coherence window" of duration T_{coh} which is then stepped across the SFT according to a given spacing criterion. All segments in each step are cross-correlated then combined incoherently.

[6] In general, for random χ^2 variables X_i, their linear combination $Y = \sum_i \mathcal{C}_i X_i$ is itself a χ^2 variable if and only if the scale coefficients $\mathcal{C}_i$ are identical (or 0). However, if the normalized coefficients $\mathcal{C}_i/\langle \mathcal{C}_i \rangle$ are close to unity, Y is reasonably *approximated* by a χ^2 distribution.

[7] **We consider specifically Type IIa GRBs as described in [94] with plateau power law decay indexes of magnitudes $\lesssim 0.5$.**

(such as those of interest to this paper) can have $N_{\rm coh}$ smaller by 1–2 orders of magnitude, so it is essential to correct the corresponding detection thresholds to account for non-Gaussianity. In particular, the Central Limit Theorem reduces the skew of the $\chi^2(2N_{\rm coh})$ distribution relatively slowly as $N_{\rm coh}$ grows.

We finally remark that, for simplicity, we have assumed coherence segments that do not overlap and no windowing function for the SFTs. For a more detailed discussion of the effects of overlapping segments and windowing, see [78].
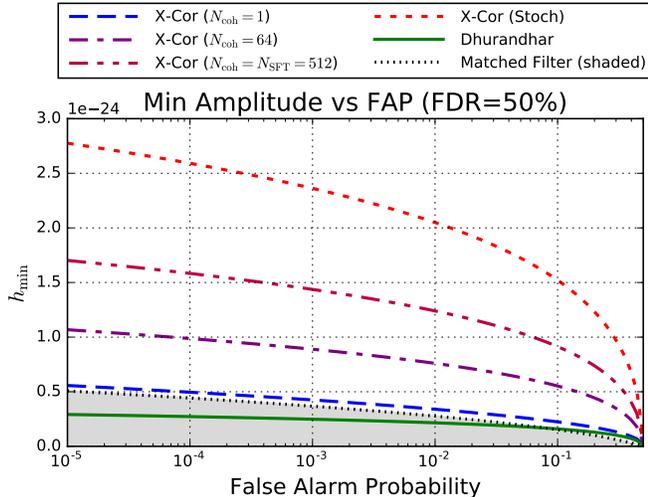
FIG. 5. **The smallest detectable GW amplitude $h_{\rm min}$ is plotted versus FAP with a set FDP of $1 - \gamma = 50\%$. A matched filter with known initial phase (black dotted line, with gray shading) is the idealized optimal search, and it provides an absolute limit on the sensitivity of *any* real search. Hence, the shaded gray area is forbidden. The matched-filter limit of the cross-correlation method (dashed blue line) is expected to approach (but not converge with) the black-dotted line. The semi-coherent limit (dash/dotted purple lines) becomes less sensitive for increasing $N_{\rm coh}$, eventually approaching the stochastic limit (dotted red). The $N_{\rm coh} = N_{\rm SFT} = 512$ limit of the cross-correlation method (red dash/double-dot red line) differs from the stochastic limit in that it includes self-pairs (autocorrelations). As discussed in the text, the assumptions of Gaussian statistics and known phase constant in [56] yield incorrect results as the resulting sensitivity (green solid line) does better than the optimal matched filter for sufficiently small FAP.**

D. Spectral Leakage Effects

Several of the assumptions made in the previous Sections are expected to lead to some amount of spectral leakage. These include the finite-time approximation of the delta function in Eq. (3.14), the quarter-cycle criterion, and SFT windowing effects (that is, the simplification of using a simple rectangular window). A full treatment of the effects of spectral leakage is outside the scope of this paper but we mention some of its effects here.

As shown in Fig. 6, spectral leakage is an issue any time the signal frequency does not precisely correspond to the center of one of the SFT frequency bins. In the simplest case of a constant frequency periodic signal, spectral leakage can cause a reduction of up to 50% in the SNR ($\mu_{\rho,\rm signal}/\sigma_{\rho,\rm noise}$) for the ρ statistic in each of the fully coherent segments. This effect is worsened when one considers time-varying frequencies: while the quarter-cycle criterion restricts the leakage from first order terms ($\dot{f}$), higher order components of the frequency evolution ($\ddot{f}$, $\dddot{f}$, etc) can lead to additional leakage. The net result is that, on average, neglecting spectral leakage will result in reduced SNR that is roughly 75% of the idealized case, see Figures 6a and 6b, and also [78].

The typical solution for this problem is to introduce a windowing function for the SFT, but this is not without tradeoffs. Each windowing function (of which there are many) has different strengths and weaknesses. The commonly used Hann window is well equipped to handle spectral leakage and maintains good frequency resolution, but suffers in amplitude accuracy [78]. SFT windows must then be overlapped in an attempt to regain some of the lost amplitude information, increasing computational cost. The Tukey window, commonly used in continuous wave searches, is – by contrast – not as good at diminishing the effects of spectral leakage but retains more of the original power. Recent work within the cross-correlation framework has examined the effects of different windowing functions in detail [77, 78].

Other methods can also be used to reduce spectral leakage. These include over-resolving each SFT by zero-padding (although this can still lead to some spectral leakage for signals whose frequency varies continuously with time), sinc-interpolating between SFT bins (thus leveraging the sampling theorem [95]), or simply adding contributions from neighboring SFT bins. Including just the two adjacent SFT bins when cross-correlating can improve recovery of the expected SNR from $\approx$77% to $\approx$90% [78].

In what follows, we acknowledge that spectral leakage could lead to SNRs that are roughly $\approx$75% of the idealized value for the ρ statistic (i.e. up to a factor of $\sqrt{.75} \approx 87\%$ in signal amplitude and/or distance reach for cases in which $\ddot{f}$ and higher terms may not be negligible). This is consistent with the estimate of 77.4% for rectangular windowing described in [78] and related searches, e.g. [77]. **Signals for which a chosen baseline is particularly close to the limit set by $\dot{f}$ via the quarter-cycle criterion (assumption #2 in Sec. III C) may experience additional leakage (not to exceed the maximum loss of $\sqrt{.5} \approx 71\%$ in**
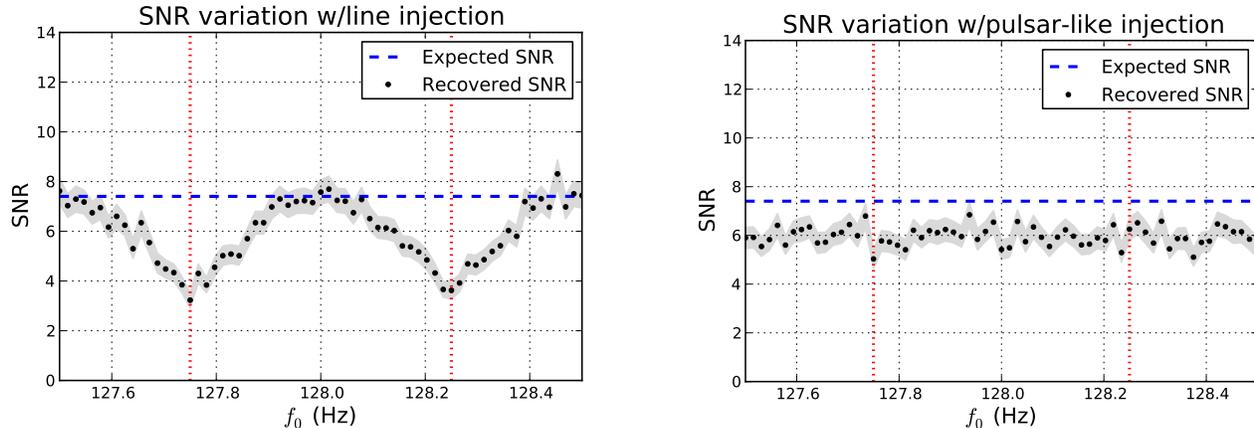
(a) SNR variation with respect to the location of f_0. The SNR for a line ($f(t) = f_0$) is at a maximum when f_0 lies at the center of one of the bins, and at a minimum when it lies on an edge. In the case of the latter, the SNR is reduced by a factor of 2, as the leakage leads to approximately half of the signal power leaking into each adjacent bin. **Thus, the maximum loss in amplitude sensitivity from spectral leakage should be no more than $\sqrt{50\%} \approx 71\%$ for a signal satisfying the quarter-cycle criterion (see Sec. III C). The grey shaded region corresponds to the 3σ error region resulting from 512 independent simulations.**

(b) SNR variation with respect to the location of f_0 for a pulsar-like evolution, $f(t) = f_0 + f_1 t$, where $f_1 = -1/1024\,\mathrm{Hz}$ was chosen such that the frequency would change by one frequency bin ($1/\Delta T_{\mathrm{SFT}}\,\mathrm{Hz}$) over the entire duration of the signal (T_{obs}), well within the quarter-cycle criterion. Here, the factor of f_1 guarantees that the signal never precisely corresponds to the center of the bin, which averages out to roughly 75% of the SNR (**or $\sqrt{75\%} \approx 87\%$ in amplitude) for any given f_0 The grey shaded region corresponds to the 3σ error region resulting from 512 independent simulations.**

FIG. 6. The effects of spectral leakage on signals of the form $h(t) = h_0 \sin \Phi(t)$ with $\Phi(t) = 2\pi \int f(t) dt$ and $h_0 = 10^{-24}$ injected into Gaussian noise with zero mean and $S_n \approx 1.75 \times 10^{-47}\,\mathrm{Hz}^{-1}$. Two frequency evolutions are considered: a line feature of constant frequency $f(t) = f_0$ (left); a pulsar-like evolution of the form $f(t) = f_0 - f_1 t$ (here $f_1 = 1/1024\,\mathrm{Hz/s}$, right). The SFT baseline used to calculate the SNR is $\Delta T_{\mathrm{SFT}} = 2$ s, resulting in SFT bin widths of $1/2$ Hz (e.g. the center of the bin located at $f_0 = 128.0$ Hz has edges at 128.25 and 127.75, red vertical dashed lines). In all cases the total duration of the signal is $T_{\mathrm{obs}} = 512$ s.

amplitude sensitivity, see Fig. 6a).

V. GRB PLATEAU SEARCH SENSITIVITY

In this Section we apply the cross-correlation statistic to the specific model of intermediate duration GW signals described in [38]. This model describes the scenario of a secularly unstable GRB-magnetar possibly associated with a GRB afterglow plateau (see also Section I).

In the Newtonian limit, the $l = m = 2$ f-mode becomes secularly unstable when the ratio $\beta = T/|W|$ of the rotational kinetic energy T to the gravitational binding energy $|W|$ is between 0.14 and 0.27. This mode has the shortest growth time of all polar fluid modes, $1\ \mathrm{s} \lesssim \tau_{GW} \lesssim 7 \times 10^4$ s for $0.24 \gtrsim \beta \gtrsim 0.15$ [54] and may be an important source of GWs. Under the hypothesis that a secular bar-mode instability does indeed set in for a magnetar left over after a GRB explosion, Corsi and Mészáros [38] have followed the NS quasi-static evolution under the effect of gravitational radiation according to the analytical formulation given by [54]. Since τ_{GW} is generally much longer than the dynamical time of the star, the evolution is quasi-static, i.e., the star evolves along an equilibrium sequence of Riemann-S ellipsoids. Differently from what was done by Lai and Shapiro [54],

Corsi and Mészáros [38] added into the evolution energy losses due to magnetic dipole radiation, assuming that those will not substantially modify the dynamics, but will act to speed up the overall evolution along the same sequence of Riemann-S ellipsoids that the NS would have followed in the absence of radiative losses.

In the model proposed by Corsi & Mészáros 2009 [38], the resulting quasi-periodic GW signal depends on five parameters: β, the initial kinetic-to-gravitational potential energy ratio of the magnetized NS [54]; n, the NS polytropic index; M, the NS mass; R_0, the unperturbed NS radius; and B_0, the initial dipolar magnetic field strength at poles. For a typical parameter choice of $M = 1.4\ \mathrm{M}_\odot$, $R_0 = 20\,\mathrm{km}$, $n = 1$, $B_0 = 10^{14}\,\mathrm{G}$, and $\beta = 0.20$ (Fig. 7, red), [38] have estimated a distance reach (assuming a matched filter search) of $\approx 100\,\mathrm{Mpc}$ for the aLIGO-Virgo detectors (for FAP$\approx 5 \times 10^{-5}$ and FDP=50%).

We have tested the detectability of this class of signals using the adaptation of the cross-correlation statistic described in the previous Sections, and assuming Gaussian noise with $S_n(f) \approx 1.83 \times 10^{-47}\,\mathrm{Hz}^{-1}$ approximately equal to that of whitened aLIGO-Virgo noise in the signal's frequency band (Fig. 2). The results are reported
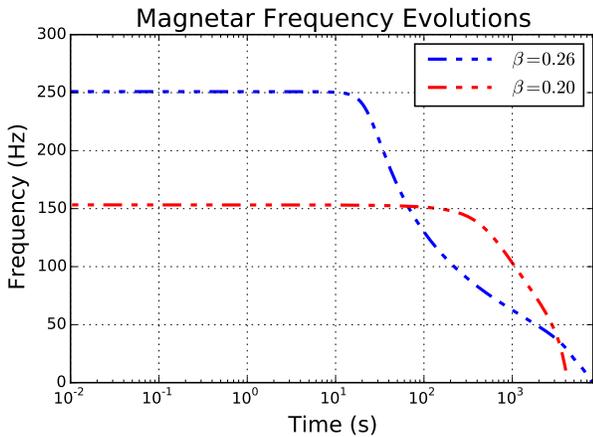
FIG. 7. Frequency evolution for two representative signals generated via the Corsi and Mészáros model. The signals are for a typical choice of parameters $M = 1.4$ M$_\odot$, $R_0 = 20$ km, $n = 1$, $B_0 = 10^{14}$ G, and different β ($\beta = 0.20$ is in the center of the allowed range of $0.14 < \beta < 0.27$). For more details on the model used to generate these waveforms, see [38].

in Table I **for an optimally oriented GRB**[8].

A matched filter analysis yields the highest sensitivity, and thus the largest horizon distance limits. For a typical choice of model parameters (e.g. $\beta = 0.20$, $B_0 = 10^{14}$ G, $M = 1.4$ M$_\odot$, $R_0 = 20$ km), if we assume that the initial phase is known as in [38], we obtain a distance limit of ≈ 139 Mpc for a FAP of 0.1% and a FDP of 50% using data from a single detector. This is consistent with the estimate of ≈ 100 Mpc reported in [38] (which assumed a smaller FAP). A real matched filter search will have an overall unknown phase constant (see the last paragraph in Sec. IV B), which reduces the horizon distance to ≈ 118 Mpc (for the same model parameters). Our cross-correlation matched filter limit yields an horizon distance of ≈ 103 Mpc, or 103 Mpc$/\sqrt{75\%} \approx 119$ Mpc when correcting for spectral leakage (see Fig. 6), in agreement with the (real) matched filter.

We finally note that signals with faster frequency evolutions are affected more by spectral leakage, for a fixed choice of $\Delta T_{\mathbf{SFT}}$ (satisfying the quarter cycle criterion). For example, the GW signal from a magnetar with $\beta = 0.26$ would have a faster frequency evolution than that from

a source with $\beta = 0.20$ (**with other source parameters unchanged; Fig. 7**). The distance horizon we achieve in the cross-correlation matched filter limit for $\beta = 0.26$ (and for a $\Delta T_{\mathbf{SFT}}$ equal to the one used for the $\beta = 0.20$ case) is ≈ 216 Mpc. This is $\approx \sqrt{57\%}$ of the expected matched filter horizon of **287 Mpc** (see Table I), worse than what we would have expected for average spectral leakage losses of $\sqrt{75\%}$, but still within the maximum expected value of $\sqrt{50\%}$ (see Fig. 6a). Such extremal losses are consistent with baselines very near to (but not exceeding) the maximum value set by the quarter-cycle criterion. These losses can be improved by optimizing the size of the baseline, given each frequency evolution, and is planned for future work. These results are summarized in **Table I.**

TABLE I. **Single-detector distance** horizons for simulations in which the search is performed on the "correct" frequency-time track for with $B_0 = 10^{14}$ G, $M = 1.4 M_\odot$, $R_0 = 20$ km and varying values of β using the model proposed by [38]. **The search techniques used are matched filtering *with unknown phase* (MF)**, the cross-correlation matched filter limit (χ^2MF, see Sec. IV B), and the cross-correlation stochastic limit (see Sec. IV A).

β Value	Distance Horizon (Mpc)		
	MF	χ^2MF	Stochastic
0.20	118	103	20
0.26	287	216	40

While a detailed study of the parameter space of the model by [38] is beyond the scope of this paper, we also carried out several simulations to demonstrate the effectiveness of a semi-coherent approach in: (i) enhancing the robustness of the search against signal uncertainties when compared to a matched-filter limit; and (ii) enhancing the sensitivity of the search when compared to a "stochastic approach". We do so by calculating the distance horizons for situations in which the *assumed* time-frequency track differs from the *actual* signal by some amount. This difference is quantified by an error (δM, δR, δB) on the values of the true signal parameters (M, R, B). The size of these errors help determine the parameter space resolution for an effective search. The results of these tests are summarized in Tables II and III.

Because an error in signal parameters implies a mismatch between the true signal time-frequency evolution and the time-frequency track adopted for the calculation of the ρ statistic, we expect the cross-correlation search to completely miss the signal in the limit of large coherence timescales, $T_{\mathrm{coh}} \to T_{\mathrm{obs}}$ (approaching the matched-filter

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[8] **Here, "optimally oriented" is taken to mean that the GRB jet is aligned with the line of sight (so that $\iota = 0$ and the GW is circularly polarized, i.e. $\mathcal{A}_+ = \mathcal{A}_\times = 1$, see Sec. III C) and the GRB sky location is such that the line of sight is orthogonal to the plane containing the detector (so that $F_+^2 + F_\times^2 = 1$).**

limit, which is not robust against such deviations). On the other hand, in the limit of small coherence timescales, $T_{\mathrm{coh}} \to \Delta T_{\mathrm{SFT}}$, while the search is expected to be robust against signal uncertainties, the sensitivity is significantly lower than the matched-filter case. Thus, for a given parameter space resolution, one can define an optimal coherence timescale, which can then be used to quantify the distance reach of the semi-coherent regime (for given FAP and FDP).

We obtain the optimal coherence time (T_{opt}) by calculating the detection efficiency for given FAP (here, 0.1%) as a function of T_{coh}, for a signal at a fixed distance. The T_{coh} that is associated with the maximal detection efficiency is then used for a series of injections of varying distance, but fixed T_{coh}. The distance that is associated with an efficiency of 50% (which is equivalent to a FDP of 50%) is then taken to be the distance horizon for that step size. The step sizes taken for each model parameter informs the size of the parameter space that a semi-coherent cross-correlation search should cover. We ran simulations with two classes of step size: "large" steps that correspond to a coarse grid in the parameter space, and "small" steps that correspond to finer (and subsequently, more computationally intensive) grid in the parameter space.

The results for the large steps are shown in Fig. 8, and summarized in Table II. Optimal coherence times, see Fig. 8 (left), are of $\mathcal{O}(1)$ s, which lead to maximal detection distances around $20 - 30$ Mpc (recovering only $\approx 25\%$ of the matched filter limit), see Fig. 8 (right). In the case where all three parameters are stepped simultaneously ($\delta \mathrm{All}$), the optimal coherence time is only twice the SFT baseline of $\Delta T_{\mathrm{SFT}} = 0.25$ s and provides no significant gains over the stochastic limit, see Table II.

TABLE II. **Single-detector distance** horizons for *large* steps in each of the model parameters with, $\beta = 0.20$, $B_0 = 10^{14}$ G $+ \delta \mathrm{B}_0$, $M = 1.4 \, M_\odot + \delta M$, $R_0 = 20$ km $+ \delta \mathrm{R}_0$. The resulting distance horizons are approximately 20-30 Mpc, which is up to a 50% improvement over the stochastic limit,[a] but only $\approx 25\%$ of the matched filter limit. All errors of order $\mathcal{O}(1)$ Mpc.

| Var | Step Size | T_{opt} (sec) | Distance Horizon (Mpc) | |
			Semi-Coh	Stochastic
δB_0	10^{12} G	1	22	20
δM	$5 \times 10^{-3} \, M_\odot$	2	28	20
δR_0	20 m	2	29	20
$\delta \mathrm{All}$	As above	0.5	20	20

[a] A factor of 1.5 in distance horizon increases the expected detection rate by a factor of $1.5^3 \approx 3$.

The small step sizes produce optimal coherence times of as high as 256 s, see Fig. 9 (left), which lead to maximal detection distances of ≈ 60-80 Mpc, roughly $\approx 75\%$

of the matched filter limit, see Fig. 9 (right). For comparison, we note that nearest long GRB on record was GRB 980425, located at a distance of 40 Mpc [96, 97]. These results suggest that the large steps considered above are indeed too large to adequately resolve the parameter space, while the small steps represent a good starting point for a finer exploration of the physically relevant parameter space. **We note that an in depth discussion of parameter space range and resolution must also include the effect of the implied number of trials on the detection statistic. This effect is expected to be more important for longer coherence times. A full study of the parameter space for intermediate-duration GWs, using the cross-correlation search technique described here, is planned for future work.**

TABLE III. **Single-detector distance** horizons for *small* steps in each of the model parameters with, $\beta = 0.20$, $B_0 = 10^{14}$ G $+ \delta \mathrm{B}_0$, $M = 1.4 \, M_\odot + \delta M$, $R_0 = 20$ km $+ \delta \mathrm{R}_0$. The simulation used $T_{\mathrm{obs}} = 1024$ s and $\Delta T_{\mathrm{SFT}} = 0.25$ s. The resulting distance horizons are approximately 60-80 Mpc, up to four times as large as the stochastic limit,[a] and $\gtrsim 75\%$ of the matched filter limit. All errors of order $\mathcal{O}(1)$ Mpc.

| Var | Step Size | T_{opt} (sec) | Distance Horizon (Mpc) | |
			Semi-Coh	Stochastic
δB_0	10^{10} G	64	61	20
δM	$5 \times 10^{-5} \, M_\odot$	256	73	20
δR_0	0.2 m	256	76	20
$\delta \mathrm{All}$	As above	64	58	20

[a] A factor of 4 in distance horizon increases the expected detection rate by a factor of $4^3 = 64$.

VI. DISCUSSION AND CONCLUSION

We have explored the application of the cross-correlation technique described in [56] to a new class of intermediate duration GW signals of duration $T_{\mathrm{obs}} \lesssim 10^4$ s, specifically the bar mode instability model for millisecond magnetars developed in [38]. In doing so, we have corrected the statistical properties of the cross-correlation statistic reported in [56] for both the semi-coherent, and full-coherent matched-filter limits. In addition, we have done a cursory exploration of the parameter space for this model.

There are several parallels between limits of the cross-correlation method and other search techniques used for LIGO data analysis. Natural examples are the techniques derived from efforts to quantify the stochastic GW background. Two such methods are the Stochastic Transient Analysis Multi-detector Pipeline (STAMP), a cross-power statistic widely used for LIGO all-sky searches [30, 98], and stochtrack, a seedless clustering algorithm

that has been tested on signal models comparable in duration to those considered here [31]. Both these methods are similar (in spirit, if not necessarily implementation) to the stochastic limit of the cross-correlation approach.

Because of their significant robustness against signal uncertainties (and relatively low computational costs) stochastic-inspired methods (as the two described above) are attractive for many search regimes, and especially as a first pass when searching for viable GW candidates with wide parameter spaces. On the other hand, the improvement in sensitivity (and therefore distance reach) enabled by the semi-coherent limit of the cross-correlation approach lends itself to deeper searches. A potential way to leverage the strengths of both regimes is to develop a framework in which a stochastic-inspired search is used for discovery, with semi-coherent cross-correlation followup for parameter estimation and refinement. This could be done entirely within cross-correlation method described in this work, or by using an established stochastic technique (e.g. STAMP) for discovery and cross-correlation for follow-up.

Overall, the results of our study are encouraging: The tunable robustness versus sensitivity of the cross-correlation technique is well suited for intermediate duration GW signals that evolve on timescales of 10^3-10^4 s, and can reach astrophysically relevant distance horizons with the expected noise characteristics of GW detectors such as aLIGO and Virgo. However, a full parameter space exploration is yet to be completed, as is testing on real instrument noise. Additionally, the trials factor for a full parameter space search will reduce, to some extent, the idealized horizon distances calculated here. We intend to explore these aspects of the analyses in future work.
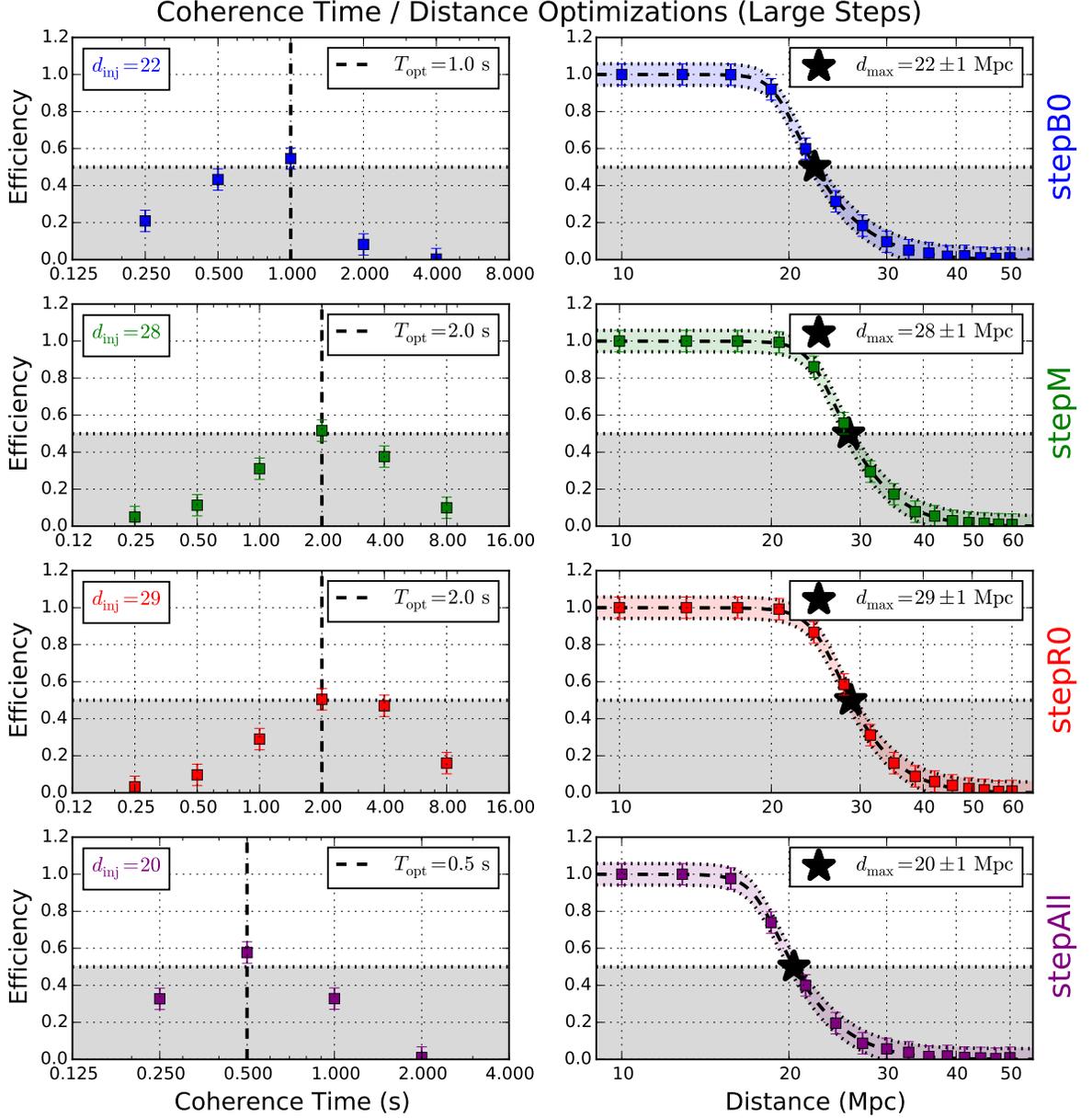
FIG. 8. Efficiency (1-FDP) plots for *large* steps $\delta B_0 = 10^{12}$ G (blue), $\delta M = 5 \times 10^{-3} M_\odot$ (green), $\delta R_0 = 20$ m (red) and all three combined (purple). All plots assume FAP=0.1% and distances are extracted using FDP=50% (black dotted line and gray shaded area). On the left, optimal coherence time plots. The signal is injected at a constant distance, $T_{\rm coh}$ is then varied to find the value that maximizes detection efficiency ($T_{\rm opt}$). On the right, $T_{\rm coh}$ is fixed at the optimum value for each step, and then distance is varied. The result is fit by an asymmetric sigmoid of the form $\text{sig(x)} = [1 + \exp(\text{p}_0\{\text{x} - \text{p}_1\})]^{-1/\text{p}_2}$ (where p_0, p_1, p_2 are constants to be fit), which is then used to interpolate and determine the max distance ($d_{\rm max}$).

FIG. 9. Efficiency (1-FDP) plots for *small* steps in $\delta B_0 = 10^{10}$ G (blue), $\delta M = 5 \times 10^{-5} M_\odot$ (green), $\delta R_0 = 0.2$ m (red) and all three combined (purple). All plots assume FAP=0.1% and distances are extracted using FDP=50% (black dotted line and gray shaded area). On the left, optimal coherence time plots. The signal is injected at a constant distance, T_{coh} is then varied to find the value that maximizes detection efficiency (T_{opt}). On the right, T_{coh} is fixed at the optimum value for each step, and then distance is varied. The result is fit by an asymmetric sigmoid of the form $\mathrm{sig(x)} = [1 + \exp(\mathrm{p_0}\{x - \mathrm{p_1}\})]^{-1/\mathrm{p_2}}$ (where $\mathrm{p_0}$, $\mathrm{p_1}$, $\mathrm{p_2}$ are constants to be fit), which is then used to interpolate and determine the max distance (d_{max}).

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