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Optimal qubit control using single flux quantum pulses

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Single flux quantum (SFQ) pulses are a natural candidate for on-chip control of superconducting qubits. We show that they can drive high-fidelity single-qubit rotations even in leaky transmon qubits, if the pulse sequence is suitably optimized. We achieve this by showing that for these restricted all-digital pulses, genetic algorithms can be made to converge to arbitrarily low error, verified up to a reduction in gate error by two orders of magnitude compared to an evenly spaced pulse train. Timing jitter of the pulses is taken into account, exploring the robustness of our optimized sequence. This takes us one step further towards on-chip qubit controls.

Introduction.—Rapid single flux quantum (RSFQ) technology has been and is originally pursued as an ultra-high-speed classical computing platform [1, 2]. The ability to generate reproducible identical pulses at high clock rate has been demonstrated in integrated circuits [3]. Next to its original motivation of ultrafast digital circuits, this makes RSFQ a highly viable candidate for the on-chip generation of control pulses and readout for quantum computers based on Josephson devices [4–9]. The switching time lies in the picoseconds range, leading to fast quantum gates [10], but the timing of the single voltage pulses to control the devices is a major challenge [11]. The integration capabilities of RSFQ are compatible with scaling quantum processors [12, 13], for example, one can load pulse-sequences into shift registers [14]. Then again, the present-day control scheme is based on room temperature electronics, whose signals are transmitted as analogue signals through filters into a cryostat, which creates large physical overhead

Besides these engineering considerations, both techniques present distinct paradigms for control design. Room temperature generators have high amplitude resolution (currently 12 bits are typical) at limited speed, thus methods of analog pulse shaping can be applied [15] to approximate the experiment with high accuracy. This has been done in superconducting qubits, e.g. within the DRAG- and WahWah method [16, 17], that suppress leakage into higher energy levels. These pulses can be readily calibrated using AdHOC [18].

On the other hand, SFQ pulses only have a single bit of amplitude resolution: in a given time interval either there is a pulse or not. The RSFQ sequence proposed in [12] again reveals the challenge to balance gate speed with suppression of leakage in transmon qubits similar to DRAG, demonstrating the need for advanced pulse design methods which cannot be solved by evenly spaced pulses alone. While their amplitude resolution is minimal, an RSFQ sequence has a very short time-constant of the pulse train, much shorter than the intrinsic frequencies of the qubit system, that can be expected to simulate analog control. To explore this, one needs to depart from the conventional optimal control paradigms as reviewed

in Ref. [15], that often involve gradients, but rather on digital control and algorithms that can optimize these fully discrete controls.

In this paper we show that optimal control methods, based on genetic algorithms adapted to digital control, can be used to improve gate operations with trains of SFQ pulses. The single bit of amplitude resolution is encoded in a discretized time evolution, which also limits possible clock frequencies. We show that uneven and relatively densely populated bit string pulse sequences can be used to suppress leakage as well as drastically reduce gate times up to the speed limit. We also verify that optimized pulse sequences are robust against timing jitter, the degree of which depends on the clock integration.

SFQ control.—As a simple model we use a qubit with a leakage level such as the transmon qubit [19, 20], that can be controlled via voltage pulses. Our single qubit target gate is a rotation around the y -axis, i.e. a Pauli- Y gate with an arbitrary global phase. The time-dependent system Hamiltonian in units of \hbar reads

$$\mathbf{H}(t) = \mathbf{H}_0 + u(t)\mathbf{H}_1, \quad (1)$$

with drift and control terms

$$\mathbf{H}_0 = \begin{bmatrix} 0 & 0 & 0 \\ 0 & \omega & 0 \\ 0 & 0 & 2\omega + \Delta \end{bmatrix}, \quad \mathbf{H}_1 = \begin{bmatrix} 0 & \frac{-i}{2} & 0 \\ \frac{i}{2} & 0 & \frac{-i}{\sqrt{2}} \\ 0 & \frac{i}{\sqrt{2}} & 0 \end{bmatrix}, \quad (2)$$

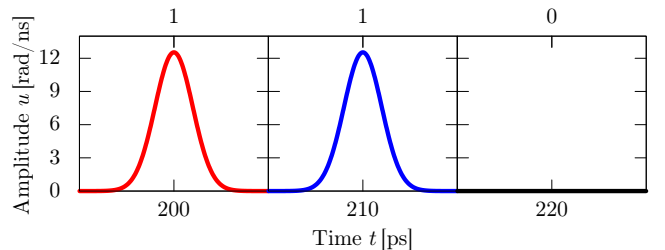


FIG. 1. (Color online) Time slicing of the evolution: in each frame the control amplitude $u(t)$ either has a Gaussian shape (1) or vanishes (0). Therefore only two unitary operators \mathbf{U}_i need to be calculated, each with a duration of $2t_c$ (here: 10 ps). The gap between two consecutive pulses is always an integer multiple of $2t_c$.

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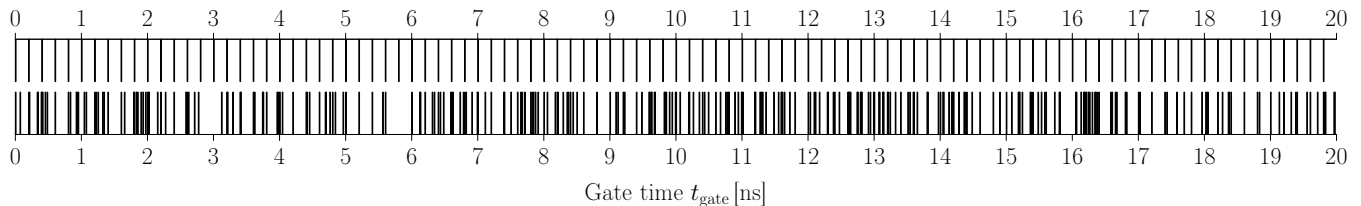


FIG. 2. Bit string representation of the pulse sequence before (top) and after (bottom) the optimization for a gate time of 20 ns. A vertical black line indicates the time when a pulse is applied, and white space the duration of the free evolution.

where ω is the qubit's angular frequency and Δ the anharmonicity of the third level. Gate operations are performed by changing the control amplitude $u(t)$. But instead of finding a proper pulse shape of duration of several nanoseconds, we apply a single picosecond pulse repeatedly. This pulse is switched on and off as stored in a shift register and controlled by a clock. The pulse shape we use is a truncated Gaussian with standard deviation τ and total duration $2t_c$

$$u(t) = \frac{\delta\theta}{\sqrt{2\pi\tau}} e^{-t^2/2\tau^2} \int_{-t_c}^{t_c} u(t) dt \approx \delta\theta. \quad (3)$$

The area $\delta\theta$ is approximated by the numerical integration of the pulse shape. It matches the rotation angle on the Bloch sphere for an infinitely narrow delta-pulse. To flip the Bloch vector around the y -axis we can thus set a lower bound of $n = \pi/\delta\theta$ pulses. Note that the three-level model outlined in Eq. (2) is sufficient to describe the qubit dynamics - we have performed *a posteriori* verifications with more levels and found no discernible difference.

We simulate our system by slicing the total gate time into N time steps of length $2t_c$, i.e. the clock period. Choosing τ accordingly ensures that the applied voltage pulse shape $u(t)$ vanishes at the beginning and the end of a the time interval. We interpret the free evolution of the system in a time interval $[t_i, t_i + 2t_c]$ as a pulse with zero amplitude. Fig. 1 shows an example of two consecutive pulses followed by an interval without an applied pulse. The time difference between two applied pulses is always a positive integer multiple of the pixel length $2t_c$. For each interval $[t_i, t_i + 2t_c]$ the time evolution is captured in a unitary matrix $\mathbf{U}(t_i) \equiv \mathbf{U}(t_i + 2t_c, t_i)$. Since we only apply a single pulse if necessary, each unitary is chosen out of a given database containing just two unitary operators, \mathbf{U}_0 and \mathbf{U}_1 . For both our work and future practical applications, this can be done efficiently by storing all the relevant system parameters in this database. Therefore the pulse sequence can be represented as bit string, where zeros represent a free evolution interval \mathbf{U}_0 and ones an applied voltage pulse \mathbf{U}_1 , see Fig. 2. The total time evolution reads

$$\mathbf{U}(t_{\text{gate}}) = \prod_{i=N-1}^0 \mathbf{U}(t_i) = \mathbf{U}(t_{N-1}) \cdots \mathbf{U}(t_0). \quad (4)$$

The database the unitaries are chosen from consists of

$$\mathbf{U}_0 = \exp(-2it_c \mathbf{H}_0) = \mathbf{U}_0(2t_c) \quad (5)$$

$$\mathbf{U}_1 = \mathbf{U}_0(t_c) \left[\mathcal{T} \exp \left(-i \int_{-t_c}^{t_c} \mathbf{H}(t) dt \right) \right] \mathbf{U}_0(-t_c). \quad (6)$$

Any adjustment in the experiment can be captured in the database and all characterization of the pulses has to be done once in order to find these two database entries.

The target gate is a Pauli-Y gate, where we allow an arbitrary phase for the leakage level, and the global phase is neglected when using an appropriate fidelity function (Eq. (8))

$$\mathbf{U}_{\text{target}} = \begin{bmatrix} 0 & -1 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & e^{i\varphi} \end{bmatrix}. \quad (7)$$

We optimize our time evolution within the computational subspace of the qubit. The average fidelity function therefore reads [21]

$$\Phi = \frac{1}{4} \left| \text{tr} \left\{ \mathbf{U}_{\text{target}}^\dagger \mathbf{P}_Q \mathbf{U}(t_{\text{gate}}) \mathbf{P}_Q \right\} \right|^2, \quad (8)$$

with the projector onto the qubit subspace \mathbf{P}_Q .

Typical values for a transmon qubit [19] are for the qubit transition frequency $\omega/2\pi = 5$ GHz and its anharmonicity $\Delta/2\pi = -200$ MHz. The pulse area and duration are set to $\delta\theta = \pi/100$ and $2t_c = 10$ ps, respectively, limiting the clock frequency to 100 GHz. The gate time chosen is $t_g = 20$ ns, leading to a total number of $N = 2000$ pixels.

As a starting point, we use the sequence presented in [12, 22], so we have exactly $n = 100$ pulses in the beginning. After each pulse we wait $t_{\text{wait}} = 2\pi/\omega - 2t_c$ for the qubit to complete a full precession before applying another pulse. This scheme increases the gate fidelities with increasing number of applied pulses, while decreasing the area underneath the pulse shape with the same rate. But it also leads to longer gate times, because every additional pulse increases the gate time by $2\pi/\omega$ due to the waiting time.

Genetic algorithms.—As gradient-based algorithms are not straightforward here, we use a genetic algorithm [23, 24] to optimize the pulse sequence. This versatile tool for global optimization is a natural candidate

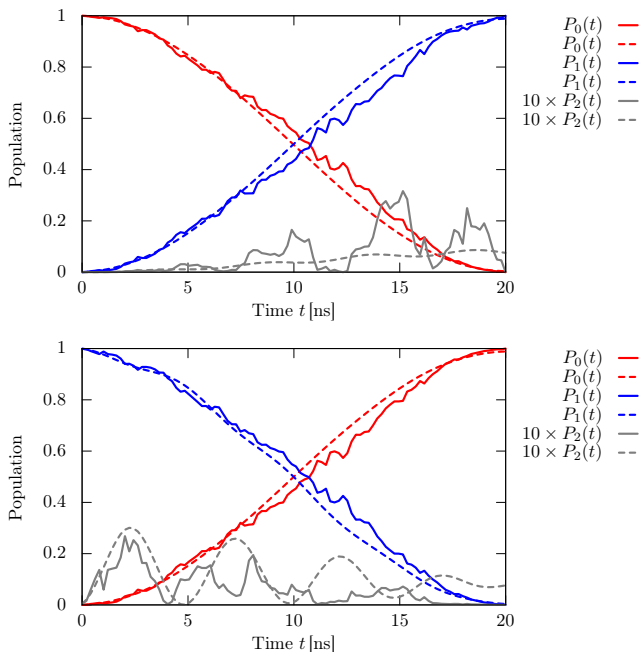


FIG. 3. (Color online) Populations of a qutrit starting in the ground (top) and excited (bottom) state for the optimized sequence (solid) and the initial sequence (dashed). The population of the leakage level is enhanced by a factor of 10 for better legibility. The algorithm suppresses leakage to the end gate time.

for the problem at hand, since our pulse sequence is already encoded into a binary string. Other genetic algorithms have been applied successfully in analogue quantum optimal control, e.g. to optimize laser pulses to control molecules [25].

Here, we search for a local minimum in the control landscape starting with the sequence described earlier [12, 22], and stop as soon as we reach the target fidelity. Within the genetic algorithm framework every solution for the variational parameters of the control problem is encoded into a genome. At each iteration a selection of genomes will be merged by crossover of genome pairs. The new and old genomes are mutated and the genomes with the best fitness make it to the next generation. Finding the right parameters for the genetic algorithm can be difficult, but it is common practice for most optimization problems to choose a high crossover and a low mutation probability [26]. The parameters of our optimization are shown in Table I.

Using that algorithm and setting our gate fidelity as fitness measure we found the solutions for the sequence shown in Fig. 2 and for the populations shown in Fig. 3. The algorithm mainly corrects for leakage into the third energy level, which always leads to an increase in the number of pulses; in the solutions presented here from $n = 100$ to $n = 301$. The time taken for a run is about 150s and the improvement is shown in Fig. 4. We encounter a broad variation over different runs indicating

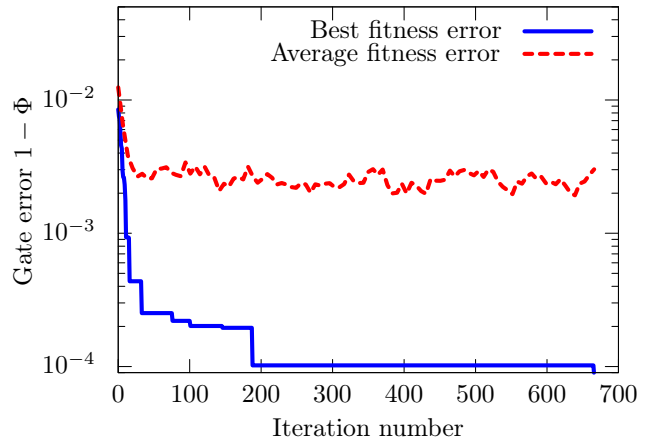


FIG. 4. (Color online) Starting with a sequence shown in [12, 22] genetic algorithms are used for optimization. At each iteration a new population of different pulse-sequences is created from their parents, while the best solution of a generation with lowest fidelity error is kept until the algorithm finds a better one or a given threshold is exceeded.

the presence of an abundance of local traps. This is consistent with the common observation that the in principle trap-free control landscape [27] develops traps when the space of available pulse shapes is strongly constrained.

Quantum speed limit.—With the genetic algorithm at hand, we can search for shorter gate times t_{gate} for the Pauli-Y gate. We keep the widths of the pulses constant, which decreases the number of pixels with a decreasing gate time. The reachable fidelities are shown in Fig. 5. The shortest possible gate time we could find within the genetic algorithm is $t_{\text{gate}} = 6$ ns. Each optimization stopped if a fidelity > 0.9999 has been reached or the maximum number of iterations (200000) was exceeded. We point out that for short gate times the evenly spaced pulse sequence is no longer a viable solution if we don't have any control of the pulse amplitude.

Timing errors.—So far the clock has been assumed to be a perfect one. Here we take inevitable timing errors into account, that lead to small pulse delays [28] and therefore to deviations from the optimal fidelity. We simulate timing errors by multiplying every applied pulse U_1 in the optimized sequence with a free evolution operator

Population size	70
Mutation probability	0.001
Crossover probability	0.9
Number of genomes to select for mating	64
Maximum allowed iterations	200000
Target fitness	0.9999
Elitism	1

TABLE I. Parameters used in the genetic algorithm. See [23] for background.

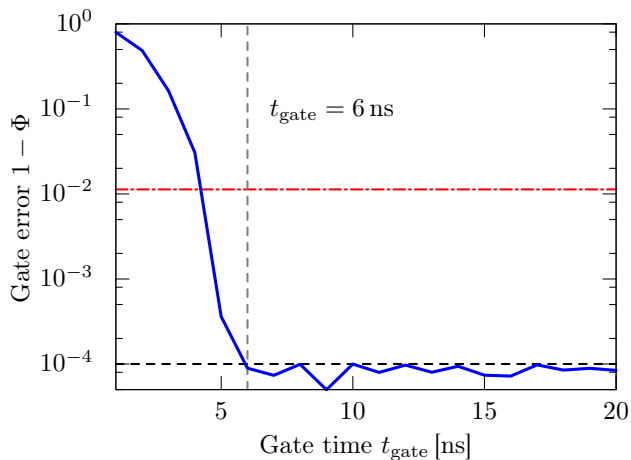


FIG. 5. (Color online) The gate errors of the optimizations for different gate times. The widths of the pulses have been kept constant, i.e. the number of pixels changes with the gate time linearly. The horizontal red dash-dotted line is the fidelity error of the initial sequence for 20 ns, and the horizontal black dashed line indicates the stopping condition of the algorithm when a fidelity > 0.9999 has been reached.

of a time interval δt from the right and its adjoint from the left. Therefore a positive δt indicates a pulse which arrives with a delay, and a negative δt indicates that the pulse arrives earlier:

$$U'_1 = U_0(-\delta t)U_1U_0(\delta t). \quad (9)$$

δt is a normal distributed random time with standard deviation σ for external jitter and $\sqrt{k}\sigma$ for internal jitter, where k is the applied pulse number. For each value of σ the fidelity of the time evolution has been averaged over 1000 runs for the optimized sequence. As can be obtained from Fig. 6, the external clocking scheme is more robust by an order of magnitude of the standard deviation. It is still within the target fidelity when the jitter time scale is 10% of the pulse width, while for an internal clock it is around 1%. If the jitter time reaches the pulse duration, the gate error is still on the same scale where we started our optimization. We therefore conclude that an external clock should be used in favor of an internal one for future devices.

Conclusion.—We have successfully developed and applied an optimal control method for pulses with only

a single bit of amplitude resolution. Finding the right binary string leads to minimization of the leakage error in the transmon system and thus gate control precision compatible with the requirements of fault tolerant quantum computing. The results presented here show a fidelity improvement of several orders of magnitude over equal pulse spacing sequences, while being robust under external timing jitter. RSFQ shift registers are needed to perform the optimized sequence and are an essential part of on-chip SFQ qubit control. This makes the underlying SFQ pulse platform together with the single

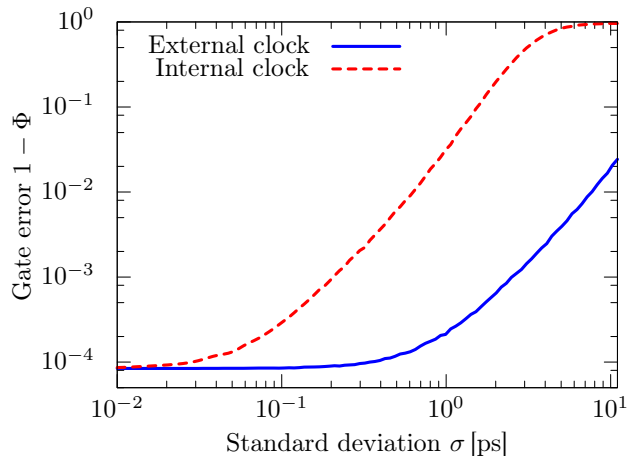


FIG. 6. (Color online) The gate error of the optimized sequence for timing jitters with constant variance (external clock) and linear growing variance (internal clock). The gate errors have been averaged over 1000 runs of the time evolution for each value of σ .

bit optimal control theory a possible and attractive candidate for an integrated control layer in a quantum processor.

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